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Full-Stokes polarimetry with circularly polarized feeds

Sources with stable linear and circular polarization in the GHz regime

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ABSTRACT

We present an analysis pipeline that allows recovering reliable information for all four Stokes parameters with high accuracy. Its novelty relies on the effective treatment of the instrumental effects already prior to the computation of the Stokes parameters contrary to conventionally used methods, such as the Müller matrix one. For instance, the instrumental linear polarization is corrected across the whole telescope beam and significant Stokes Q and U can be recovered even when the recorded signals are severely corrupted by instrumental effects. The accuracy we reach in terms of polarization degree is of the order of 0.1–0.2 %. The polarization angles are determined with an accuracy of almost 1°. The presented methodology was applied to recover the radio linear and circular polarization of around 150 Active Galactic Nuclei. The sources were monitored between July 2010 and April 2016 with the Effelsberg 100-m telescope at 4.85 GHz and 8.35 GHz with a cadence of around 1.2 months. The polarized emission of the Moon was used to calibrate the polarization angle of the monitored sources. Our analysis showed a small system-induced rotation of about 1° at both observing frequencies. Finally, we find five sources with significant and stable linear polarization; three sources remain constantly linearly unpolarized over the period we examined; a total of 11 sources have stable circular polarization degree m_c and four of them with non-zero m_c . We also identify eight sources that maintain a stable polarization angle over the examined period. All this is provided to the community for future polarization observations reference. We finally show that our analysis method is conceptually different from the traditionally used ones and performs better than the Müller matrix method. Although it has been developed for a system equipped with circularly polarized feeds it can easily be generalized to systems with linearly polarized feeds as well.

1. Introduction

As an intrinsic property of non-thermal emission mechanisms, the linear and circular polarization from astrophysical sources carry information about the physical conditions and processes in the radiating regions (e.g. Laing 1980; Wardle et al. 1998; Homan et al. 2009). Propagation through birefringent material – such as the intergalactic and interstellar medium – can further generate, modify, or even eliminate the polarized part of the transmitted radiation (e.g. Pacholczyk 1970; Jones & O’Dell 1977; Huang & Shcherbakov 2011). Consequently, processes that introduce variability in the emitting or transmitting regions induce dynamics in the observed polarization parameters (e.g. Marscher et al. 2008; Myserlis et al. 2014). Although these processes increase the complexity, they carry information about the mechanisms operating at the emitting regions.

The measured degree of linear and especially circular polarization of extragalactic sources in the radio window, is usually remarkably low. Klein et al. (2003) studied the B3-VLA sample in the range from 2 GHz to 10 GHz to find that the average linear polarization degree ranges from $\sim 3.5\%$ to 5% . Myserlis (2015) studied the circular polarization of almost 45 blazars in the GHz regime and found a population median of around 0.4% . Consequently, despite its importance, the reliable detection of polarized emission is particularly challenging especially when propagation and instrumental effects as well as variability processes are considered.

In the following, we present a pipeline for the reconstruction of the linear and circular polarization parameters of radio sources. The pipeline includes several correction steps to minimize the effect of instrumental polarization, allowing the detection of linear and circular polarization degrees as low as 0.3% . The instrumental linear polarization is calculated across the whole telescope beam and hence it can be corrected for the observations of both point-like and extended sources. The methodology was developed for the 4.85 GHz and 8.35 GHz receivers of the 100 m Effelsberg telescope. Although these systems are equipped with circularly polarized feeds, our approach can be easily generalized for telescopes with linearly polarized feeds, as well.

The consistency of our method is tested with the study of the most stable sources in our sample (in terms of both linear and circular polarization). Their stability indicates that physical conditions such as the ordering, magnitude or orientation of their magnetic field remain unchanged over long timescales. The corresponding polarization parameters are reported for the calibration of polarization observations. We report both polarized and randomly polarized (unpolarized) sources. Conventionally, the latter are used for estimating the instrumental effects and the former to calibrate the data sets and quantify their variability.

The paper is structured as follows. In Section 2 an introduction to the technical aspects of the observations is presented. In Section 3 we present the methodology we developed to extract the polarization parameters with high accuracy. Our approach

relies mainly on the careful treatment of the instrumental linear and circular polarization, discussed in Sect. 3.2 and 3.7, respectively, as well as the correction of instrumental rotation presented in Sect. 3.6. In Section 4 we perform a qualitative comparison between our method and the Müller matrix one. In Section 5 we describe the statistical analysis of the results obtained with our methodology and report on sources with stable linear and circular polarization. Finally, a discussion and the conclusions of our work are presented in Section 6.

Throughout the manuscript we use the conventions adopted by Commissions 25 and 40 at the 15th General Assembly of the IAU in 1973:

- a) the reference frame of Stokes parameters Q and U is that of right ascension and declination with the polarization angle starting from north and increasing through east, and
- b) positive circular polarization measurements correspond to right handed circular polarization,

This circular polarization convention is also in agreement with the Institute of Electrical and Electronics Engineers (IEEE) standard, according to which the electric field of a positive or right handed circularly polarized electromagnetic wave rotates clockwise for an observer looking in the direction of propagation (IEEE Standards Board 1979).

2. Observations

The data set we discuss here was obtained with the Effelsberg 100 m telescope at 4.85 GHz and 8.35 GHz. The corresponding receivers are equipped with circularly polarized feeds (Table 1). A short description of the Stokes parameter measurement process using such systems is provided in Sect. 3.1.

The data set covers the period between July 2010 and April 2016. Until January 2015 the observations were conducted within the framework of the F-GAMMA monitoring program¹ (Fuhrmann et al. 2016); beyond January 2015, data were obtained as part of multi-frequency monitoring campaigns on selected sources. The median cadence is around 1.2 months. The average duration of the observing sessions is 1.3 days.

Table 1: Receiver characteristics.

Receiver		4.85 GHz	8.35 GHz
Bandwidth	(GHz)	0.5	1.1
System Temperature	(K)	27	22
FWHM	(arcsec)	146	81
Number of feeds		2	1
Polarization		LCP, RCP	LCP, RCP
Sensitivity	(K/Jy)	1.55	1.35

The observations were conducted with “cross-scans”, that is by slewing the telescope beam over the source position in two perpendicular directions. For the data set considered here those passes (hereafter termed “sub-scans”) were performed along the azimuth and elevation directions. The advantage of the cross-scan method is that it allows correcting for the power loss caused by possible telescope pointing offsets. A detailed description of the observing technique is given by Angelakis et al. (2015).

In Sect. 3 we give a detailed description of the methodology followed for the reconstruction of the total flux density I ,

the degree of linear and circular polarization m_l and m_c , and the polarization angle, χ . The number of sources with at least one significant measurement (signal-to-noise ratio, $\text{SNR} \geq 3$) of any of I , m_l , χ or m_c , are shown in Table 2, where we also list the mean uncertainty and cadence of the corresponding data sets at both frequencies.

Table 2: Number of sources with at least one significant data point ($\text{SNR} \geq 3$), mean uncertainty and cadence for I , m_l , m_c , and χ .

Data set	Units	4.85 GHz	8.35 GHz
I	sources	155	150
	uncertainty (mJy)	19	15
	cadence (months)	1.3	1.3
m_l	sources	90	96
	uncertainty (%)	0.1	0.1
χ	cadence (months)	1.7	1.7
	sources	90	96
	uncertainty ($^\circ$)	1	0.6
m_c	cadence (months)	1.7	1.7
	sources	63	54
	uncertainty (%)	0.1	0.1
	cadence (months)	6	12

3. Full-Stokes polarimetry

In the current section we present the steps taken for reconstructing the circular and linear polarization parameters of the incident radiation from the observables delivered by the telescope. Our approach aims at recovering the polarization state outside the terrestrial atmosphere.

Our methodology is readily applicable to systems with circularly polarized feeds and it can be easily modified for systems with linearly polarized feeds. The latter are sensitive to the horizontal, $E_h(t)$, and vertical, $E_v(t)$, linearly polarized electric field components of the incident radiation. The Stokes parameters in terms of these components can be written as:

$$I = \langle E_h^*(t)E_h(t) \rangle + \langle E_v^*(t)E_v(t) \rangle = \langle E_H^2 \rangle + \langle E_V^2 \rangle, \quad (1)$$

$$Q = \langle E_h^*(t)E_h(t) \rangle - \langle E_v^*(t)E_v(t) \rangle = \langle E_H^2 \rangle - \langle E_V^2 \rangle, \quad (2)$$

$$U = 2\text{Re}(\langle E_h^*(t)E_v(t) \rangle) = 2 \langle E_H E_V \cos \delta \rangle, \quad (3)$$

$$V = 2\text{Im}(\langle E_h^*(t)E_v(t) \rangle) = 2 \langle E_H E_V \sin \delta \rangle, \quad (4)$$

where,

$E_{H,V}$ the amplitudes of the two orthogonal linearly polarized electric field components

δ the phase difference between $E_h(t)$ and $E_v(t)$.

A comparison between the Stokes parameterizations for linear (Eqs. 1–4) and circular bases (Eqs. 7–10) shows that for systems with linearly polarized feeds the treatment of Stokes I needs not to be changed while Stokes Q , U and V should be treated as Stokes V , Q and U for systems with circularly polarized feeds, respectively. As an example, for systems with linearly polarized feeds, the instrumental polarization correction scheme presented in Sect. 3.2 should be applied in the U - V instead of the Q - U space, while the analysis of Sect. 3.7 should be applied to Stokes Q instead of V .

¹ <http://www.mpifr-bonn.mpg.de/div/vlbi/fgamma>

Figure 1 serves as a schematic summary of the analysis sequence. Each analysis level is labeled with an index (e.g. L1, L2 etc.) and is discussed in the section noted in that flow chart. The mean effect of each correction step is listed in Table 3.

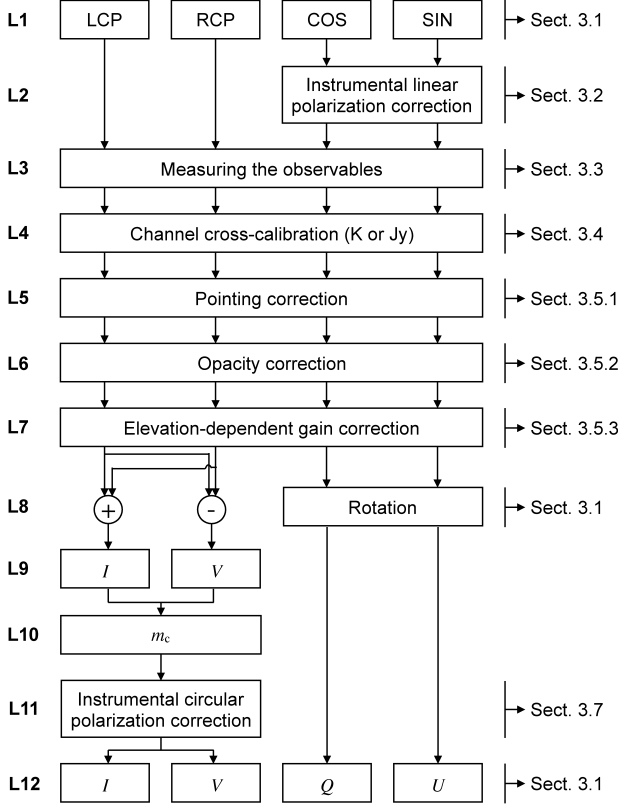


Fig. 1: A schematic summary of the analysis sequence. Each analysis level is labeled with an index on the left and is discussed in the section noted on the right. The mean effect of each correction step is listed in Table 3.

Table 3: Average percentage effect of each correction step on all Stokes parameters. Left columns refer to 4.85 GHz while right ones to 8.35 GHz.

	I		Q		U		V	
Instrumental LP	≤ 0.5	≤ 0.5	≤ 0.5	≤ 0.5
Pointing	0.4	0.9	0.4	0.9	0.4	0.8	3.6	1.4
Opacity	4.0	4.0	4.0	4.0	4.0	4.0	4.0	4.0
Gain curve	1.0	0.9	1.0	0.9	1.0	0.9	1.0	0.9
Instrumental CP	0.0	0.0	89.0	118.5

3.1. Measuring the Stokes parameters

Because our receivers are sensitive to the left- and right-hand circularly polarized components of the electric field, it is convenient to express the incident radiation in a circular basis:

$$E_L(t) = E_L e^{i\omega t}, \quad (5)$$

$$E_R(t) = E_R e^{i(\omega t + \delta)} \quad (6)$$

where,

- $E_{L,R}$ the amplitudes of the two (orthogonal) circularly polarized electric field components
- ω the angular frequency of the electromagnetic wave
- δ the phase difference between $E_L(t)$ and $E_R(t)$.

The four Stokes parameters can then be written in terms of $E_L(t)$ and $E_R(t)$ (omitting the impedance factors), as:

$$I = \langle E_L^2 \rangle + \langle E_R^2 \rangle, \quad (7)$$

$$Q = 2 \langle E_L E_R \cos \delta \rangle, \quad (8)$$

$$U = 2 \langle E_L E_R \sin \delta \rangle = 2 \langle E_L E_R \cos(\delta - 90^\circ) \rangle, \quad (9)$$

$$V = \langle E_R^2 \rangle - \langle E_L^2 \rangle, \quad (10)$$

where $\langle \rangle$ denotes averaging over time to eliminate random temporal fluctuations of E_L , E_R and δ (e.g. Cohen 1958; Kraus 1966). A detailed discussion of the Stokes parametrization is given by Chandrasekhar (1950); Kraus (1966); Jackson (1998).

The system measures the four Stokes parameters by correlation operations, i.e. multiplication and time averaging of the signals $E_L(t)$ and $E_R(t)$, based on the parametrization of Eqs. 7–10:

$$I = \langle E_L^*(t) E_L(t) \rangle + \langle E_R^*(t) E_R(t) \rangle, \quad (11)$$

$$Q = 2 \langle E_L^*(t) E_R(t) \rangle, \quad (12)$$

$$U = 2 \langle E_L^*(t) E_R(t) \rangle_{90^\circ}, \quad (13)$$

$$V = \langle E_R^*(t) E_R(t) \rangle - \langle E_L^*(t) E_L(t) \rangle, \quad (14)$$

where the “*” denotes the complex conjugate and the subscript “90°” of Eq. 13 denotes that the correlation is performed after an additional phase difference of 90° is introduced between $E_L(t)$ and $E_R(t)$. The auto-correlations of $E_L(t)$ and $E_R(t)$ – needed for I (Eq. 11) and V (Eq. 14) – are processed separately in two receiver channels labeled LCP and RCP, respectively. On the other hand, the two cross-correlations of $E_L(t)$ and $E_R(t)$ – needed for Q (Eq. 12) and U (Eq. 13) – are delivered in yet another pair of channels labeled COS and SIN. The LCP, RCP, COS and SIN channel data sets constitute the input for our pipeline (Fig. 1, level L1).

The alt-azimuthal mounting of the telescope introduces a rotation of the polarization vector in the Q - U plane by the parallactic angle, q (Fig. 1, level L8). In the general case, a potential gain difference between the COS and SIN channels introduces an additional rotation, ϕ . The angle ϕ vanishes once we balance the COS and SIN channel gains (Sect. 3.4; Fig. 1, level L4) but we need to take it into account when we calculate Stokes Q and U using the COS and SIN signals before the channel cross-calibration:

$$\begin{bmatrix} Q \\ U \end{bmatrix} = \begin{bmatrix} \cos(2q + \phi) & -\sin(2q + \phi) \\ \sin(2q + \phi) & \cos(2q + \phi) \end{bmatrix} \cdot \begin{bmatrix} \text{SIN} \\ -\text{COS} \end{bmatrix} \quad (15)$$

Throughout the following analysis we occasionally express Q and U in either of the default north-east or the azimuth-elevation (azi-elv) reference frames. We differentiate the latter case by explicitly using the notation $Q_{\text{azi,elv}}$ or $U_{\text{azi,elv}}$, which can be calculated by setting $q = 0$ in Eq. 15.

In Fig. 2 we show an example of an ~11% linearly polarized point-like source in LCP, RCP, COS and SIN channels. The Stokes I , Q , U and V are computed from Eqs. 7–10 and 15 once the source amplitude in those channels is known (Sect. 3.3; Fig. 1, level L12).

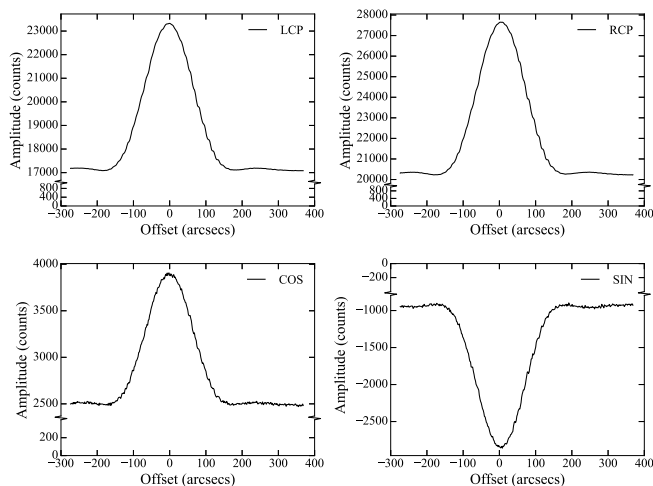


Fig. 2: An example of the same sub-scan in all channels on the source 3C 286 at 4.85 GHz. The abscissa is the offset from the commanded position of the source. The LCP and RCP channel data sets are shown in the upper row while the COS and SIN data sets in the lower row. The target source is $\sim 11\%$ linearly polarized.

3.1.1. Feed ellipticity and the measurement of Stokes parameters

Equations 5 to 15 describe the measurement of Stokes parameters for systems with ideal circularly polarized feeds. For such systems, the recorded left- and right-hand circularly polarized electric field components are perfectly orthogonal. In reality, instrumental imperfections lead to a (slight) ellipticity of the circular feed response. In this case, a small fraction of the incident left-hand circularly polarized electric field component is recorded by the right-hand circularly polarized channel of the system and vice versa.

The feed ellipticity can lead to deviations of the measured Stokes parameters from the incident ones. In Appendix B we provide an elementary approach to derive a rough estimate of the effect for the systems we used. A thorough study instead can be found in e.g. McKinnon (1992) or Cenacchi et al. (2009). For Stokes I , we estimate that those deviations are at the level of 1 mJy for our dataset, which is much less than the average uncertainty of our measurements (15–20 mJy, Table 2). Stokes Q and U on the other hand can be significantly modified by the feed ellipticity effect. A novel methodology to correct for the system-induced linear polarization across the whole telescope beam is described in Sect. 3.2. Finally, Stokes V is practically not affected in that simplified approach because the additional LCP and RCP terms introduced by the feed ellipticity (Eqs. B.15 and B.16) cancel out. Nevertheless, as described in Sect. 3.7 our measurements suffer from instrumental circular polarization, which is most likely caused by a gain imbalance between the LCP and RCP channels. Two independent methodologies for the instrumental circular polarization correction are presented in Sect. 3.7.1.

3.2. Correcting for instrumental linear polarization

Instrumental imperfections manifest themselves as a cross-talk between the signals $E_l(t)$ and $E_r(t)$. This can be best seen in unpolarized sources for which their cross-correlation is not null

contrary to what is theoretically expected (Eq. 8 and 9). Figure 3 shows an example of $Q_{azi,elv}$ and $U_{azi,elv}$ data sets of a linearly unpolarized source. Instead of the expected constant, noise-like signal, spurious patterns are clearly visible. Their amplitudes can be up to $\sim 0.5\%$ of the total flux density (Fig. 4). Those signals can also be interpreted as “slices” of the polarized beam patterns over the azimuth and elevation directions.

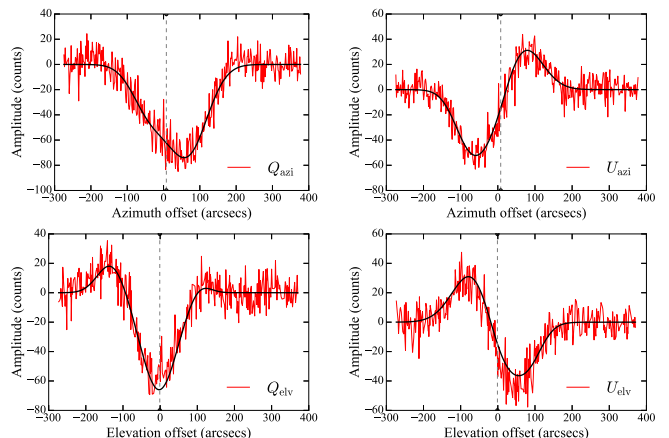


Fig. 3: Stokes $Q_{azi,elv}$ (left column) and $U_{azi,elv}$ (right column) data sets recorded at 4.85 GHz for the unpolarized, point-like source NGC 7027 in the two scanning directions, the azimuth in the top and the elevation in the bottom row. Instead of the expected flat, noise-like pattern, the presence of spurious signals are clearly visible in both scanning directions. The instrumental polarization, calculated by the instrument model M , is shown in each panel with a smooth black line. The dashed grey lines mark the source position along the scanning direction.

To correct for the instrumental polarization (Fig. 1, level L2), we describe the telescope response to unpolarized sources with one instrument model M for each of the $Q_{azi,elv}$ and $U_{azi,elv}$ along azimuth and elevation. Each M is written as a sum of j Gaussians or first derivatives of Gaussians selected empirically:

$$M = \sum_{j=1}^{\leq 3} F_j(\alpha_j I, (\mu - \beta_j), \gamma_j \sigma) \quad (16)$$

where, F_j is a Gaussian or first derivative of Gaussian with amplitude $\alpha_j I$, peak offset $(\mu - \beta_j)$ and full width at half maximum (FWHM) $\gamma_j \sigma$. The explicit functional form of the models we used for the 4.85 GHz and 8.35 GHz receivers are given in Appendix A. The parameters I , μ and σ in Eq. 16 are the measured mean amplitude, peak offset and FWHM in the LCP and RCP channels, respectively. The identification of the optimal instrument model for a given observing session from this family of models requires the evaluation of the parameters α_j , β_j and γ_j .

For the evaluation of α_j , β_j and γ_j we fit all observations on linearly unpolarized sources simultaneously. We first concatenate all sub-scans (index i in Eq. 17) on all sources (index k). Subsequently, each of the four data sets is fitted with a function of the form:

$$A = \sum_k \sum_i M_{ik} \quad (17)$$

In these terms, A is simply a concatenation of a total of $i \cdot k$ instrument models of the form M . In Fig. 4 we show the fitted instrument models for 65 observing sessions. The variability of

the plotted models is comparable to the respective errors of the fit, which indicates that the instrumental polarization remained fairly stable throughout the period of 5.5 years we examined.

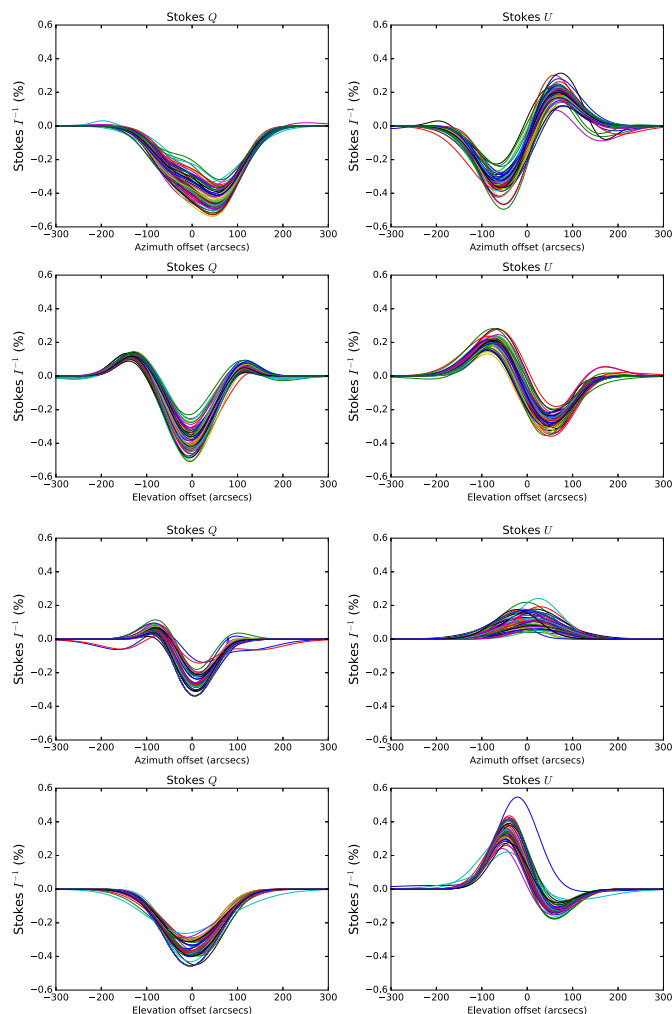


Fig. 4: The generated Stokes Q and U instrument models in the two scanning directions, the azimuth in the top and the elevation in the bottom panels for (a) the 4.85 GHz (upper two rows) and (b) 8.35 GHz (lower two rows) receivers. The models were generated for 65 observing sessions. The variability of the plotted models is comparable to the respective errors of the fit.

After having optimized α_j , β_j and γ_j for a given session, we remove the instrumental polarization from each sub-scan, in two steps:

1. we first substitute the measured I , μ and σ in Eq. 16 to determine the explicit form of the instrumental polarization in that sub-scan;
2. we then subtract this instrumental effect from the observed $Q_{\text{azi,elv}}$ and $U_{\text{azi,elv}}$.

An example is shown in Fig. 5. With this approach, the scatter of the measured polarization parameters can be dramatically decreased owing to the fact that each sub-scan is treated separately (Fig. 6).

3.3. Measuring the observables

Once $Q_{\text{azi,elv}}$ and $U_{\text{azi,elv}}$ have been corrected for instrumental polarization the process of measuring the Stokes parameters requires first the precise determination of the source amplitudes in the LCP, RCP, COS and SIN channels (Fig. 1, level L3). As an example, given the low degree of circular polarization m_c expected for our sources and because V is the difference between their amplitudes in LCP and RCP (Eq. 10), an accuracy of at least 0.1 % to 0.3 % is required for an uncertainty of no more than about 0.1 % to 0.2 % in the m_c . This precision would correspond to a 3–5 σ significance for a 0.5 % circularly polarized source.

Our tests showed that the most essential element for the amplitude measurement is the accurate knowledge of the telescope response pattern and particularly the accurate determination of the baseline level. We found that the antenna pattern for a uniformly illuminated circular aperture which is described by the Airy disk function:

$$I = I_0 \left[\frac{2J_1(x)}{x} \right]^2, \quad (18)$$

delivers significantly more accurate results than the commonly used Gaussian function, mainly because the latter fails to provide a precise description of the response beyond the FWHM. In Eq. 18, I_0 is the maximum response level of the pattern at the center of the main lobe and J_1 is the Bessel function of the first kind. In reality, the Effelsberg 100m telescope beam is described by a more complex expression since its aperture is not uniformly illuminated, mainly due to the supporting structure of the secondary reflector. Nevertheless, the amplitude uncertainties using the Airy disk antenna pattern approximation (0.1 %–0.2 %) are small enough to accommodate reliable low circular polarization degree measurements. Figure 7 demonstrates the effectiveness of the Gaussian and the Airy disk beam pattern models in terms of the fractional residuals when we fit the observed data. Those are clearly minimized in the case of the Airy disk pattern.

Ideally the Airy disk could also be used for modeling the instrumental polarization (Eq. 16) instead of Gaussians. This however would cause only an insignificant improvement (a small fraction of a percent) in the knowledge of the instrumental polarization magnitude. It would require a several-hundred-Jy source to cause a measurable effect.

3.4. Channel cross-calibration

Because of inevitable gain differences between LCP, RCP, COS and SIN receiver channels, their response to the same photon influx is generally different (different number of “counts”). The level balancing – cross-calibration – of their signals is necessary before accurate polarization measurements can be conducted (Fig. 1, level L4).

The cross-calibration is performed by the periodical injection of a known-polarization signal at the feed point of the receiver every 64 ms with a duration of 32 ms. For the used receivers this signal is generated by a noise diode designed to be: (a) circularly unpolarized, and (b) completely linearly polarized at a given polarization angle. The noise diode amplitude for each channel can be estimated as the average difference of the telescope response with the noise diode “on” and “off” in that channel. The cross-channel calibration is then achieved by expressing the LCP, RCP, COS and SIN amplitudes in noise diode units. The noise diode signal can be further calibrated to physical units, e.g. Jy, by com-

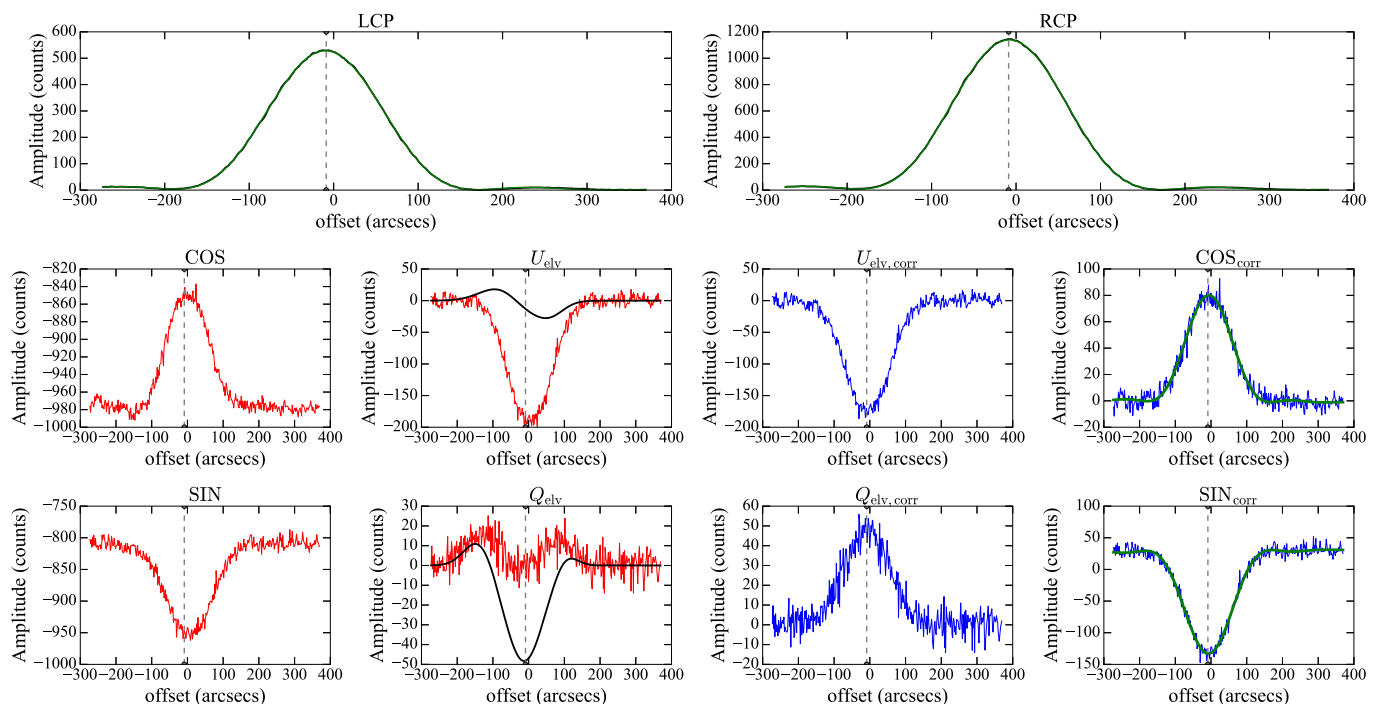


Fig. 5: Example of the instrumental linear polarization correction for an elevation sub-scan at 4.85 GHz. The observed COS/SIN (or equivalently $Q_{\text{elv}}/U_{\text{elv}}$) signals are shown in the middle and bottom rows, right below the LCP signal (with red), while the corrected signals are shown right below the RCP signal (with blue). The correction is performed by subtracting the expected instrumental polarization signals (smooth black lines) from the observed Q_{elv} and U_{elv} data sets (2nd and 3rd column of the middle and bottom rows). The instrumental polarization signals are calculated by substituting the measured mean amplitude, I , peak offset, x and FWHM, σ , of the LCP and RCP signals (top row) in the instrument model, M , created for the given observing session.

parison with reference sources (e.g. Ott et al. 1994; Baars et al. 1977; Zijlstra et al. 2008).

3.5. Post-measurement Corrections

Before the final calculation of the Stokes parameters, the source amplitudes in channels LCP, RCP, COS and SIN are subjected to a list of post-measurement corrections which are discussed in detail in Angelakis (2007); Angelakis et al. (2009, 2015); Myserlis (2015).

3.5.1. Pointing correction

This step corrects for the power loss caused by offsets between the true source position and the cross-section of the two scanning directions (Fig. 1, level L5). Imperfect pointing may potentially also increase the scatter of the amplitudes from different sub-scans as, in the general case, the offset depends on the scanning direction. Assuming an Airy disk beam pattern the amplitude corrected for a pointing offset p_{off} , is

$$A_{\text{poi}}^{\text{azi,elv}} = A^{\text{azi,elv}} \cdot \left(\frac{2J_1(w \cdot p_{\text{off}}^{\text{elv,azi}})}{w \cdot p_{\text{off}}^{\text{elv,azi}}} \right)^{-2} \quad (19)$$

where,

- azi, elv denotes the scanning direction,
- A_{poi} the source amplitude corrected for pointing offset,
- A the uncorrected source amplitude,
- p_{off} the average absolute offset in arcsecs on the other scanning direction,
- w is calculated as

$$w = \frac{3.23266}{\text{FWHM}} \quad (20)$$

where FWHM is the full width at half maximum of the telescope at the observing frequency in arcsecs.

It is important to note that the offset in the one direction (e.g. elevation) is used for correcting the amplitude in the other direction (e.g. azimuth).

The pointing correction is performed independently for each channel as the beam patterns are generally separated on the plane of the sky due to the miss-alignment between the feeds and the main axis of the telescope (“beam-squint” effect, e.g. Heiles 2002). In Fig. 8, we plot the density contours of all measured offsets from the source position separately for each channel. The beam-squint is directly evident as the miss-alignment of the contour peaks. For polarimetric observations, the beam-squint can introduce fake circular polarization, since the LCP and RCP beam patterns measure the source with different sensitivities.

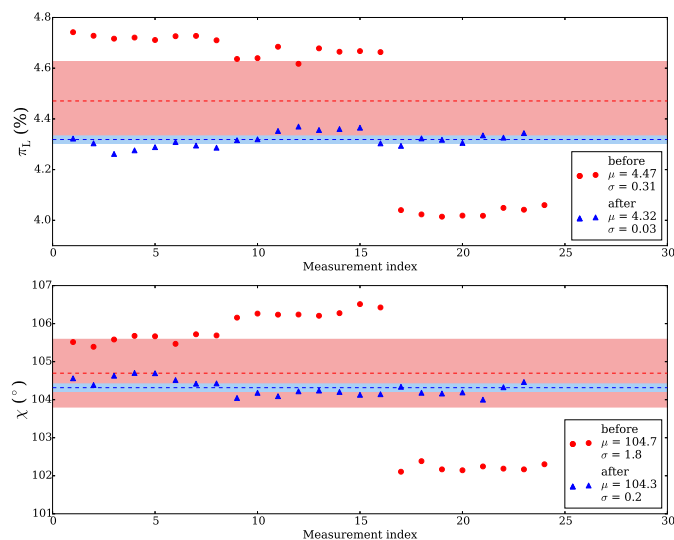


Fig. 6: Degree of linear polarization (top panel) and polarization angle (bottom panel) of the source 3C 48 at 4.85 GHz before (red circles) and after (blue triangles) applying the instrumental linear polarization correction. The data correspond to 24 sub-scans of the source within a single observing session. The average values are marked with dotted lines and the highlighted areas indicate the $1\text{-}\sigma$ region around them. The respective values are shown in the legend.

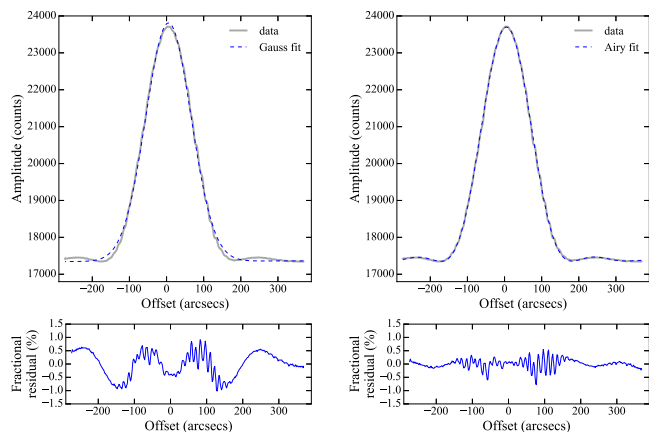


Fig. 7: A comparison between the Gaussian (left column) and Airy disk (right column) beam pattern models. The recorded data sets are shown in the top row with thick, grey lines and the model fits with dashed, blue lines. The corresponding fractional residuals, $(data - model)/data$, are given in the bottom row for a direct comparison between the models.

3.5.2. Opacity correction

The opacity correction corrects for the signal attenuation caused by the Earth's atmosphere and relies on the calculation of the atmospheric opacity at the source position, τ_{atm} (Fig. 1, level L6).

Given the amplitude A of a measurement at an elevation ELV , the amplitude corrected for atmospheric opacity will be

$$A_{\text{opc}} = A \cdot e^{\tau_{\text{atm}}} \quad (21)$$

where, τ_{atm} the atmospheric opacity at ELV . Under the assumption of a simple atmosphere model, τ_{atm} can be computed as a

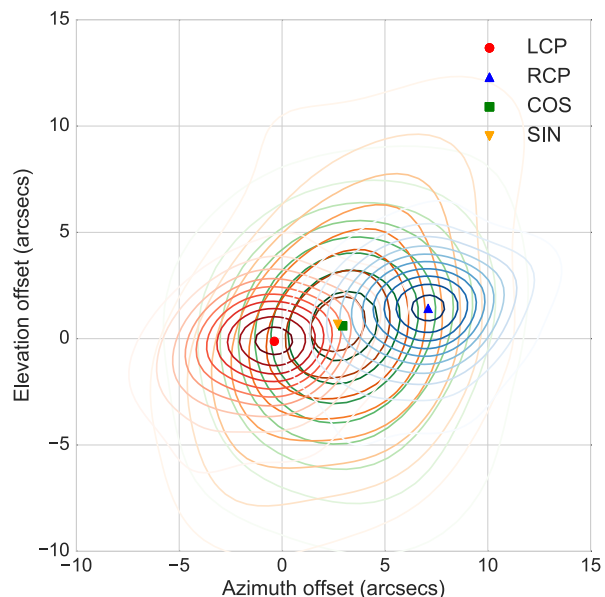


Fig. 8: Density plots of the measured offsets between the source position and the LCP, RCP, COS and SIN beam patterns for the 4.85 GHz receiver. The beam-squint can be clearly seen by the miss-alignment of the contour peaks (LCP: circle, RCP: upward-looking triangle, COS: square, SIN: downward-looking triangle).

simple function of the zenith opacity τ_z

$$\tau_{\text{atm}} = \tau_z \cdot AM = \tau_z \frac{1}{\sin(ELV)} \quad (22)$$

with AM the airmass at ELV .

For any given observing session, a linear lower envelope is fitted to the airmass (AM) - system temperature (T_{sys}) scatter plot (Fig. 9) with T_{sys} taken from the off-source segment of all sub-scans. It can be shown that the inferred slope is a direct measure of the atmospheric opacity at the zenith, τ_z (Angelakis 2007; Angelakis et al. 2009).

As we show in the example session of Fig. 9, τ_z is independent of LCP and RCP channels implying that the atmospheric absorption does not influence the polarization of the transmitted radiation. Hence, we applied the same τ_{atm} values to correct the amplitudes in all channels.

3.5.3. Elevation-dependent gain correction

The last correction accounts for the dependence of the telescope gain on elevation caused by the gravitational deformation of the telescope's surface (Fig. 1, level L7).

The amplitude corrected for this effect given a value A measured at elevation ELV , will be

$$A_{\text{gc}} = \frac{A}{G(ELV)} \quad (23)$$

where, $G(ELV)$ the gain at elevation ELV . The gain is assumed to be a second order polynomial function of ELV . The parameters of the parabola used here have been taken from the Effelsberg website².

² <https://eff100mwiki.mpifr-bonn.mpg.de>

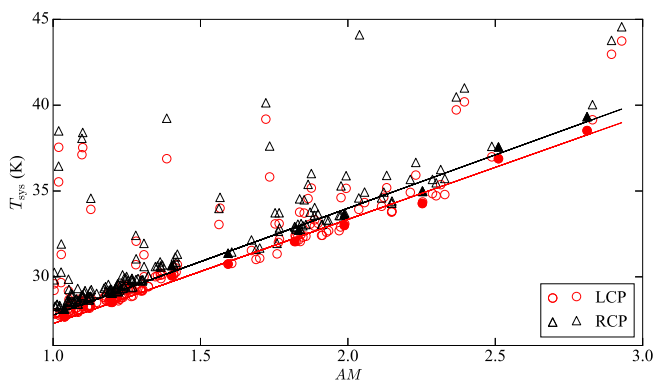


Fig. 9: LCP and RCP system temperature (T_{sys}) versus the air-mass (AM) for one observing session at 4.85 GHz. The AM range is split in a given number of bins and the data points with the lowest T_{sys} within each bin (filled markers) are used to fit the two lower envelopes (solid lines). Their slopes are practically identical, indicating that the atmospheric absorption does not influence the polarization of the transmitted radiation.

Figure 10 shows the gain curve computed for one selected session. As it is seen there the least-square-fit parabolas for the LCP and RCP data sets are very similar. Hence, the gravitational deformations do not affect the polarization measurements giving us the freedom to use the same correction factors G for all channels.

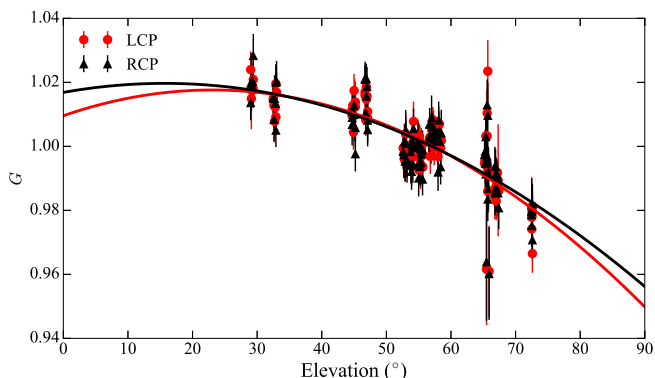


Fig. 10: Elevation - Gain curves separately for LCP and RCP data sets at 4.85 GHz. The similarity of the fitted parabolas indicate that the gravitational deformations do not affect the polarization measurements.

As it was already stated, the correction factors used for the opacity and elevation-dependent gain corrections were identical for all channels. Consequently, they do not affect fractional expressions of the Stokes parameters such as the polarization degree or angle (Eqs. 40, 41 and 42). Yet they are included formally in this step of the analysis as they affect the values of the total and polarized flux densities.

3.6. Correcting for instrumental rotation

Imprecise knowledge of the noise diode polarization angle potentially leads to poor knowledge of the power to be expected in

the COS and SIN channels. Consequently, this leads to imperfect channel cross-calibration which will manifest itself as an instrumental rotation. To study this effect we conducted observations of the Moon which has a stable and well understood configuration of the polarization orientation.

The lunar black body radiation is linearly polarized. The polarization degree maximizes close to the limb while the polarization angle has an almost perfect radial configuration (e.g. Heiles & Drake 1963; Poppi et al. 2002; Perley & Butler 2013b).

We first performed the usual azimuth and elevation cross-scans centered on the Moon. Before estimating the polarization angle – which was the objective of this exercise – the observed $Q_{\text{azi,elv}}$ and $U_{\text{azi,elv}}$ were corrected for instrumental polarization. As we discuss in Sect. 3.2 the instrument model for a sub-scan on a point source depends on the parameters measured in that sub-scan: I , μ and σ . For extended sources the brightness distribution is needed instead. For this reason, for each sub-scan we first recovered the lunar brightness distribution by de-convolving the observed I with an Airy disk beam pattern. The de-convolution is then used to evaluate the explicit form of the instrumental polarization for that sub-scan by convolving the corresponding instrument model M with the calculated brightness distribution. $Q_{\text{azi,elv}}$ and $U_{\text{azi,elv}}$ were then corrected for the instrumental polarization, which was calculated across the whole extend of the source.

We restricted the comparison of the observed and the expected polarization angle at the four points of the Moon’s limb that we probed. That is north, south, east and west. To quantify the instrumental rotation, we compared the median polarization angle around those four limb points with the values expected for the radial configuration. The east and west limb points are expected to be at 90° (or -90°), while the north and south limbs at 0° (or 180°). On the basis of 62 measurements at 4.85 GHz and 40 at 8.35 GHz, our analysis yields an average offset of $1.26^\circ \pm 0.11^\circ$ for the former and $-0.50^\circ \pm 0.12^\circ$ for the latter. These are the values we consider the best-guess for the instrumental rotation. All polarization angles reported in this paper have been corrected for this rotation.

There is evidence that the instrumental rotation depends on elevation. A Spearman’s test over all points on the Moon’s limb yielded a ρ of 0.69 ($p = 0.01$) and 0.79 ($p = 0.02$) for the 4.85 GHz and 8.35 GHz, respectively. Aside from this being a low-significance result it is also based on a narrow elevation range ($\sim 32.5^\circ$ – 50°). Yet, it is an indication that the instrumental rotation may have a more complex behavior.

3.7. Correcting for instrumental circular polarization

Imbalances between LCP and RCP channels similar to the ones discussed in Sect. 3.6 for COS and SIN, can introduce instrumental circular polarization. Two effects with which the instrumental circular polarization is manifested are:

1. As we show in Fig. 11, the distributions of the circular polarization degree, m_c , measurements are centered around a non-zero value.
2. We measure systematically non-zero circular polarization from circularly unpolarized sources. An example is the case of the planetary nebula NGC 7027 (point-like at the two frequencies we consider), a free-free emitter expected to be circularly and linearly unpolarized, for which non-zero circular polarization is measured (Fig. 11).

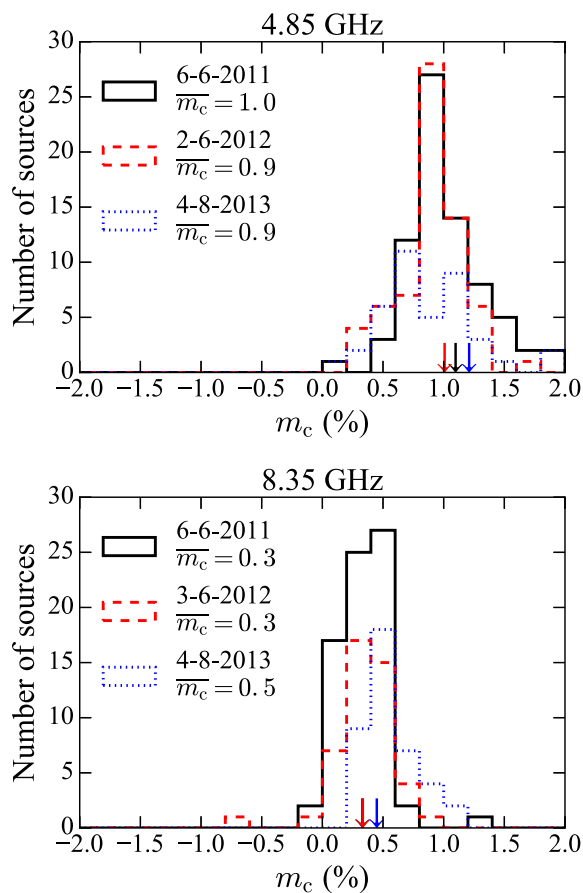


Fig. 11: Distributions of the circular polarization degree, m_c , measurements at 4.85 (top) and 8.35 GHz (bottom) for three observing sessions. The non-zero average values, shown in the legend, indicate the presence of instrumental circular polarization. The m_c measurements of the planetary nebula NGC 7027, in the corresponding sessions, are marked with arrows (in the bottom plot, two arrows overlap).

The instrumental circular polarization may be as high as $\sim 0.5\%$ to 1% – comparable to the mean population values at those frequencies – and shows significant variability. The latter is indicated by the significantly correlated, concurrent variability we observe in m_c light curves of different sources. For example, in Fig. 12, we plot the locally normalized discrete correlation function (DCF, [Lehar et al. 1992](#); [Edelson & Krolik 1988](#)) between the m_c light curves of two randomly chosen bright sources in our sample, namely 4C+38.41 and CTA 102 before and after the instrumental polarization correction. For the uncorrected data, the most prominent maxima appear at zero time lag where the correlation factor is 0.8 ± 0.2 and 0.7 ± 0.2 for the 4.85 GHz and 8.35 GHz data, respectively. These are also the only DCF maxima above the 3σ significance level.

3.7.1. Correction methods

If L_D and R_D is the diode signal in the LCP and RCP channels, the measured amplitudes of a source i in a session j expressed in

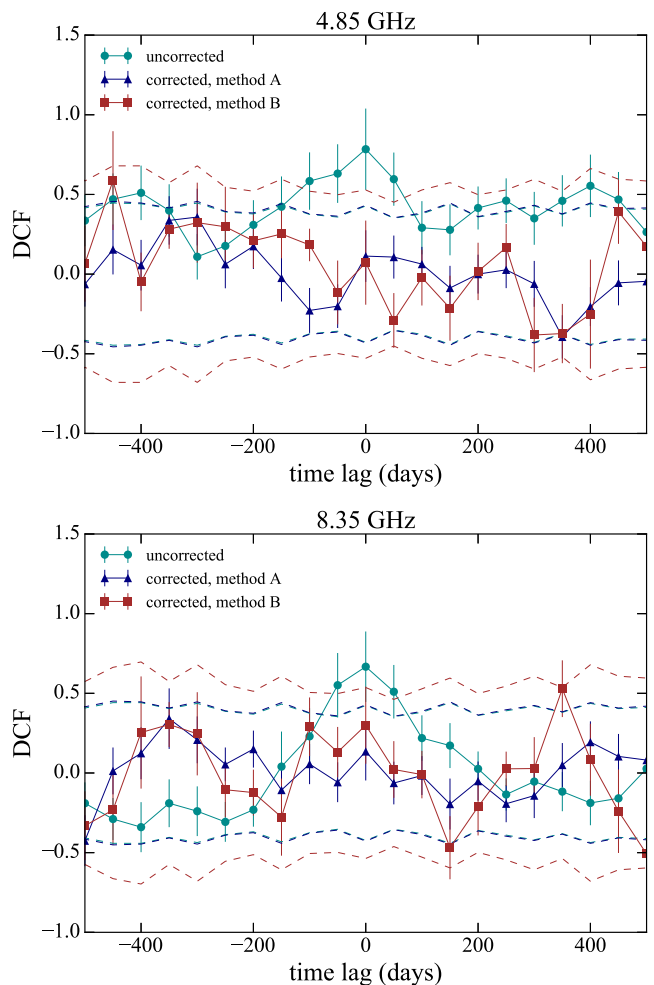


Fig. 12: Discrete correlation function (DCF) between the m_c light curves of 4C+38.41 and CTA 102 before (cyan circles) and after the instrumental circular polarization correction using either method A (blue triangles) or method B (red squares) as described in the text. The corresponding 3σ significance levels are shown with dashed lines.

diode units, will be:

$$L'_{ij} = \frac{L_{ij}}{L_{D,j}} \quad \text{and} \quad (24)$$

$$R'_{ij} = \frac{R_{ij}}{R_{D,j}}, \quad (25)$$

where L_{ij} and R_{ij} are the source incident signals modulated only by the channel gain imbalance. The degree of circular polarization of the incident radiation can be recovered from the measured amplitudes L'_{ij} and R'_{ij} by estimating the ratio $r = R_D/L_D$. Under the assumption that the noise diode is truly circularly unpolarized r becomes unity. In reality that is not the case and instrumental circular polarization emerges.

Using Eqs. 7, 10, 24 and 25, the corrected circular polarization degree can be written as:

$$m_{c,ij} = \frac{R_{ij} - L_{ij}}{L_{ij} + R_{ij}} = \frac{m'_{c,ij}r_j + m'_{c,ij} + r_j - 1}{m'_{c,ij}r_j - m'_{c,ij} + r_j + 1}, \quad (26)$$

where,

$$m'_{c,ij} \text{ the measured circular polarization degree estimated using } L'_{ij} \text{ and } R'_{ij}$$

$$m'_{c,ij} = \frac{R'_{ij} - L'_{ij}}{L'_{ij} + R'_{ij}}. \quad (27)$$

Thus, in order to recover the corrected circular polarization degree $m_{c,ij}$, we need to determine the ratio r_j . In the following we show two independent methods to compute it (Fig. 1, level L11).

Method A: Zero-level of m_c

The first method relies on the determination of the circular polarization degree $m'_{c,ij}$ that we would measure if the incident radiation was circularly unpolarized (zero level), i.e. $m_{c,ij} = 0$. We consider two estimates of the zero level:

1. the circular polarization degree of unpolarized sources (e.g. NGC 7027), and
2. the average circular polarization degree of a sufficiently large, unbiased collection of sources.

We then compute r_j by using either of these measures as $m'_{c,ij}$ in Eq. 26 and setting $m_{c,ij} = 0$. To avoid biases caused by small number statistics, we used the second measure only for sessions where at least 20 sources were observed. To make sure that the average circular polarization degree is not affected by sources which are significantly polarized, we applied an iterative process to exclude them from the calculation similar to the *gain transfer* technique presented in Homan et al. (2001) and Homan & Lister (2006).

As shown in Fig. 13, the r_j values calculated by either of the two measures are in excellent agreement. Depending on the data availability, the one or the other measure was used. For the subset of sessions where both measures were available the average r_j was used.

Method B: Singular value decomposition (SVD)

The second method requires the presence of a number of stable circular polarization sources within our sample, independently of whether they are polarized or not. If we divide Eqs. 24 and 25 we get

$$\frac{L'_{ij}}{R'_{ij}} = r_j \frac{L_{ij}}{R_{ij}} \Rightarrow Q'_{ij} = r_j Q_{ij} \quad (28)$$

For a source i with constant circular polarization, Q_{ij} does not depend on the session j , so that $Q_{ij} = Q_i$. Consequently, any variability seen in the measured Q'_{ij} for these sources can only be induced by the system through r_j . For the sources of constant circular polarization it can be written

$$Q'_{ij} = Q_i r_j \quad (29)$$

or in matrix-vector form

$$\mathbf{Q}' = \mathbf{q} \mathbf{r}^T. \quad (30)$$

Using the singular value decomposition method (SVD, e.g. Golub & Van Loan 2013), we can express matrix \mathbf{Q}' as a sum

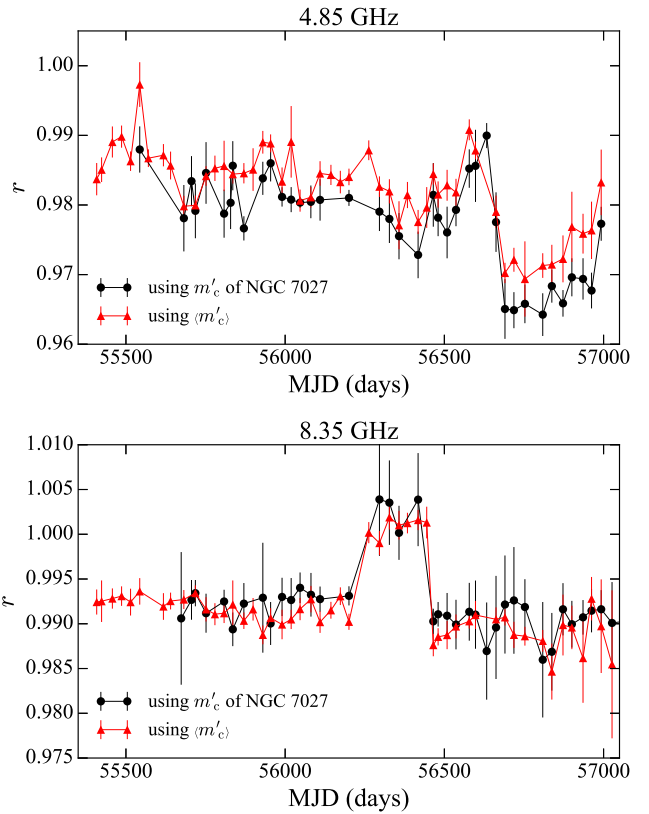


Fig. 13: The ratio r estimated using both measures of the circular polarization zero-level as described in method A.

of matrices each of which has rank one. For $i = 1, \dots, n$ sources with stable circular polarization observed over $j = 1, \dots, m$ sessions and $n \leq m$, it will be

$$\mathbf{Q}' = \sigma_1 \mathbf{u}_1 \mathbf{v}_1^H + \dots + \sigma_n \mathbf{u}_n \mathbf{v}_n^H, \quad (31)$$

where σ_i are the singular values of the matrix \mathbf{Q}' in decreasing magnitude and \mathbf{u}_i and \mathbf{v}_i are its left- and right-singular vectors, respectively. \mathbf{u}_i is of length n and \mathbf{v}_i of length m . If $\sigma_1/\sigma_2 \gg 1$, \mathbf{Q}' can be approximated by the first term in Eq. 31 and one can write,

$$\mathbf{Q}' = \mathbf{q} \mathbf{r}^T \simeq \sigma_1 \mathbf{u}_1 \mathbf{v}_1^H. \quad (32)$$

This implies that the unknown vectors \mathbf{q} and \mathbf{r} are parallel to \mathbf{u}_1 and \mathbf{v}_1 , respectively:

$$\mathbf{q} = a \mathbf{u}_1 \quad (33)$$

$$\mathbf{r} = b \mathbf{v}_1 \quad (34)$$

with $ab = \sigma_1$. In order to solve for a and b , we need at least one source of known circular polarization included in the list of n stable sources. Assuming that this is the first source of the set ($i = 1$), Q_1 can be calculated from

$$Q_1 = \frac{1 - m_{c,1}}{m_{c,1} + 1} \quad (35)$$

where $m_{c,1}$ is its known circular polarization degree, and using Eq. 29 we can write

$$\sum_j Q'_{1j} = \sum_j Q_1 r_j = Q_1 \sum_j r_j \quad (36)$$

Additionally, from Eq. 34, we have

$$\sum_j r_j = \sum_j bv_j \Big|_1 = b \sum_j v_j \Big|_1 \quad (37)$$

where $\sum_j v_j \Big|_1$ is the sum of all the elements of vector \mathbf{v}_1 . Therefore, using Eqs. 36 and 37, the factor b can be computed as:

$$b = \frac{\sum_j r_j}{\sum_j v_j \Big|_1} = \frac{\sum_j Q'_{1j}}{Q_1 \sum_j v_j \Big|_1} \quad (38)$$

and the factor $a = \sigma_1/b$. Once b has been computed, Eq. 34 will give us vector \mathbf{r} , the elements of which are the circular polarization correction factors r_j to be used in Eq. 26.

The SVD methodology was implemented using three sources: NGC 7027, 3C 48 and 3C 286. This subset of sources was selected as the best candidates with stable circular polarization based on the following criteria:

1. Stability of the observed circular polarization (even being unpolarized). Assuming that the observed variability is a superposition of the instrumental and the intrinsic polarization variability, the sources with the lowest m'_c variability are the best candidates to be intrinsically stable.
2. More frequently observed. This criterion ensures that we can apply Method B to as many sessions as possible and account for the instrumental polarization that can show pronounced variability even in short timescales (Fig. 13).

For the analyzed datasets, the sources 3C 286 and 3C 48 best fulfill both of the above criteria. NGC 7027 was selected as the source assumed to have known circular polarization. Its free-free emission is expected to be circularly unpolarized ($m_c = 0$) and hence its $Q = 1$ according to Eq. 35. For a session with no NGC 7027 data, we adopted a mock source of zero circular polarization to which we assigned as observed value the m'_c averaged over all sources in that session. A minimum of 20 sources was required in those cases. The circular polarization degree values for the other sources assumed stable (3C 286 and 3C 48 in our case) are not required for Method B, which is one of the main advantages of this calibration technique.

The circular polarization stability of the selected sources is also advocated by:

1. The fact that they display the most significantly correlated, concurrent variability in circular polarization (Fig. 14). Their low circular polarization variability is supported by the coincidence of all lines there, as well as by the fact that for them the zero time lag correlation exceeds the 5σ threshold.
2. The σ_1/σ_2 ratios for the 4.85 GHz and 8.35 GHz data using these sources were ~ 521 and ~ 540 (~ 27 dB), respectively justifying the approximation of Q'_{ij} with a single rank-one matrix.

3.7.2. Comparison between methods A, B and the UMRAO database

In Fig. 12 we show the DCF of corrected circular polarization data for 4C+38.41 and CTA 102. The data corrected with methods A and B are shown separately. The improvement is directly evident in the radical decrease of the correlation factors at zero time lag. In fact, the zero time lag correlation does not exceed the 1σ level.

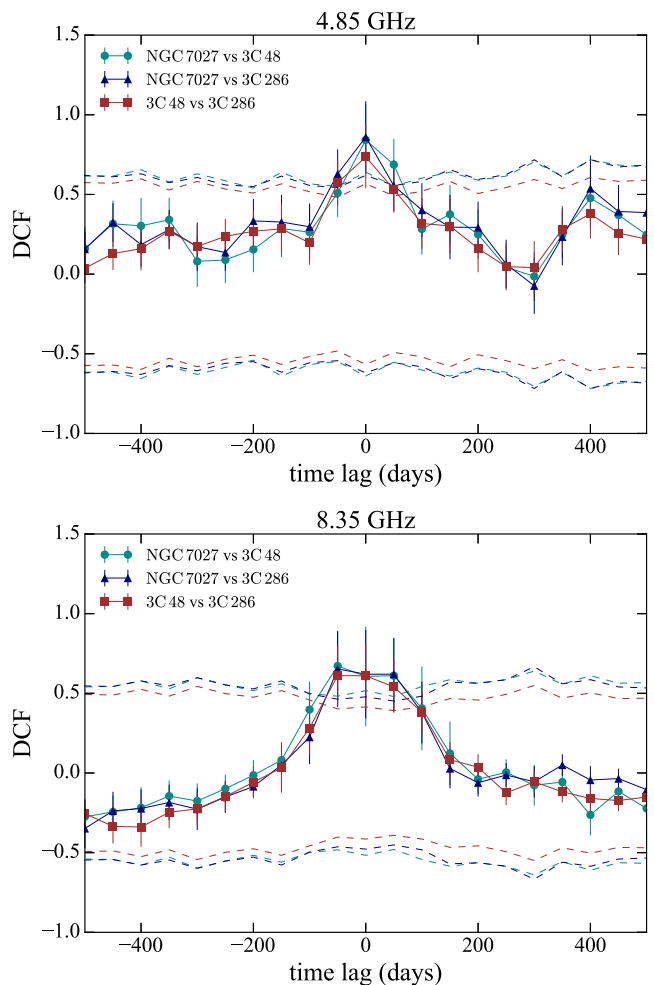


Fig. 14: Discrete correlation function (DCF) between the m'_c light curves of all source pairs including NGC 7027 and 3C 48 and 3C 286. The most prominent maxima above the the 5σ significance levels, indicated by the corresponding dashed lines, appear around zero time lag where the correlation factor is $\sim 0.8 \pm 0.2$ and $\sim 0.6 \pm 0.2$ for the 4.85 GHz and 8.35 GHz data, respectively. The dashed lines indicate the 5σ significance level.

Figure 15 now shows the m_c measurements of NGC 7027 before and after the correction for instrumental circular polarization with methods A and B. For the uncorrected data sets, we systematically measure non-zero m_c with standard deviations of 0.4 % and 0.2 % at 4.85 GHz and 8.35 GHz, respectively. The corrected m_c on the other hand, from methods A and B: are in excellent agreement; appear very close to zero; and their standard deviations are reduced to 0.1 %.

For the sources with the most stable behavior, Table 9 lists their circular polarization measurements. Methods A and B agree well within the errors. Method B gives on average 0.016 % smaller standard deviations than method A does. This is most likely caused because method B assumes more sources with constant polarization than method A.

Finally, we compared the circular polarization measurements from method A and B with measurements from the UMRAO monitoring program (Aller & Aller 2013; Aller et al. 2016). The comparison was performed for five sources with overlapping

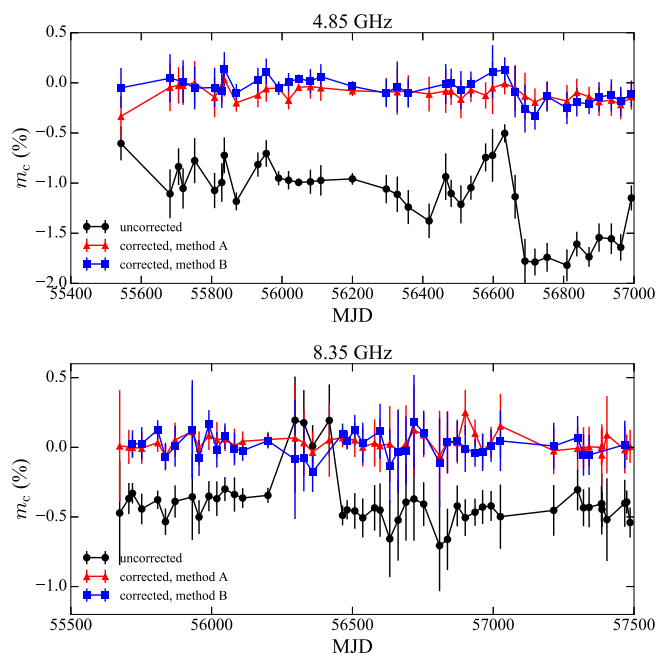


Fig. 15: The m_c measurements of NGC 7027 at 4.85 GHz and 8.35 GHz before and after the instrumental circular polarization correction using both methods A and B.

data sets from both monitoring programs (~ 2010.5 – 2012.3), namely 3C 84, OJ 287, 3C 279, BL Lacertae and 3C 454.3. Specifically, we compared all concurrent data points within a maximum separation of 2 weeks. There are 110 and 59 such data points of the UMRAO data set overlapping with the results of methods A and B, respectively. In both cases we found a median absolute difference in the circular polarization degree measurements of only 0.2 %. The corresponding data sets for the time range embracing the overlapping period are shown in Fig. C.1. The comparison of such contemporaneous measurements is particularly important because it can be used to detect or put strict limits on the very rapid variations usually observed in circular polarization.

4. Comparison with the Müller matrix method

Traditionally the Müller matrix method has been the one adopted for treating the instrumental polarization. Here we wish to carry out a comparison with our method.

The Müller matrix method is based on estimating the elements of the Müller matrix \mathbf{M} , which is a transfer function between the incident S_{real} and the measured S_{obs} Stokes 4-vectors:

$$S_{\text{obs}} = \mathbf{M} \cdot S_{\text{real}} \quad (39)$$

A set of four independent measurements of sources with known S_{real} are enough to solve the system of equations 39 and compute the \mathbf{M} matrix elements. In case more measurements are available a fit can determine the best-guess values. The inverse \mathbf{M} matrix is then applied to S_{obs} to correct for instrumental effects.

As we discussed in Sect. 3.2, our methodology models and corrects for the instrumental linear polarization across the whole beam before extracting the Stokes Q and U data, while the Müller method has no handle on this. This can be essential for

cases of low Q or U amplitudes which can be corrupted to the point that the telescope response pattern cannot be seen in the data. For the majority of such cases our methodology was able to recover the telescope response pattern (e.g. Q_{eliv} in Fig. 5). In milder cases our methodology resulted in peak offsets closer to the source position and FWHM values closer to the actual ones.

Another advantage of our approach is the milder conditions it requires. As we discussed in Sect. 3.7.1, the only requirement for method A is the observation of a circularly unpolarized source and for method B n sources with constant circular polarization with the need to know the exact value of m_c only for one of them. The Müller method on the other hand, requires a good coverage of the Stokes parameter space and particularly for V .

To perform a quantitative comparison of the two techniques, we focused on the linear polarization results. First, we calculated I , Q and U for all sources and observing sessions using both our methodology and the Müller method, accounting for all post-measurement corrections described in Sect. 3.5. For each session, the 3x3 Müller matrix was determined using all observations on the polarization calibrators shown in Table 4.

Table 4: The polarization calibrators used for the Müller matrix method. Their parameters were provided by Dr. A. Kraus (priv.comm.).

Source	Freq. (GHz)	I (Jy)	m_1 (%)	χ ($^\circ$)
3C 286	4.85	7.48	11.19	33.0
	8.35	5.22	11.19	33.0
3C 295	4.85	6.56	0.00	0.0
	8.35	3.47	0.93	28.9
3C 48	4.85	5.48	4.19	106.6
	8.35	3.25	5.39	114.5
NGC 7027	4.85	5.48	0.00	0.0
	8.35	5.92	0.00	0.0

As a figure of merit for the comparison, we used the intra-session variability of I , Q and U in terms of their standard deviation $\sigma_{I,Q,U}$. Since our sources are not expected to vary within a session (not longer than 3 days), any variability in I , Q and U can naturally be attributed to instrumental effects. Consequently, the technique leading to lower variability must be providing a better handling of the instrumental effects. In our study we included measurements with linearly polarized flux of at least 15 mJy.

We performed two-sample Kolmogorov-Smirnov (KS) tests to compare the corresponding $\sigma_{I,Q,U}$ distributions between the two techniques for three polarized flux ranges: all, high polarization (≥ 100 mJy), low polarization (15 mJy – 100 mJy). The results for the 4.85 GHz receiver are presented in Table 5 along with the median $\sigma_{I,Q,U}$ values for either of the two techniques. The cases where the KS test rejects the null hypothesis that the two distributions are the same at a level greater than 5σ are marked with an asterisk.

For Stokes I , both methods perform equally well since the corresponding KS-test results show no significant difference between the σ_I distributions of the Müller method and our methodology. That is also the case for Stokes Q and U of the high polarization data.

Our approach performs significantly better for Stokes Q and U when we consider either the complete dataset (all) or the low polarization data. The corresponding KS-test results show that the $\sigma_{Q,U}$ distributions of the Müller method and our methodology are significantly different above the level of 5σ . A direct

comparison of the median $\sigma_{Q,U}$ values shows that our method delivers $\sim 8\%$ and $\sim 28\%$ more stable results for the complete dataset (all) and the low polarization data, respectively. The main reason for this improvement is the instrumental linear polarization correction scheme of our methodology that treats each sub-scan separately, accounting for the instrumental polarization contribution across the whole beam (Sect. 3.2).

5. Sources with stable polarization

The methodology described in Sect. 3 was used to compute the linear and circular polarization parameters of the observed sources at 4.85 GHz and 8.35 GHz. Once all four Stokes parameters have been computed, the degree of linear and circular polarization, m_l and m_c and the polarization angle χ , were calculated as:

$$m_l = \frac{\sqrt{Q^2 + U^2}}{I}, \quad (40)$$

$$m_c = \frac{V}{I}, \quad (41)$$

$$\chi = \frac{1}{2} \arctan \frac{U}{Q}. \quad (42)$$

The corresponding errors were computed as the Gaussian error propagation of the uncertainties in the LCP, RCP, COS, and SIN amplitudes through Eqs. 7–10, 15 and 40–42. Finally, for each observing session, we computed the weighted average and standard deviation of the polarization parameters for all sub-scans on a given source, using their errors as weights.

Our data set includes a total of 155 sources and was searched for cases of stable linear and circular polarization characteristics to be listed as reference sources for future polarization observations. Because we are interested in identifying only cases with stable polarization parameters, we restricted our search to a subsample of 64 sources that were observed:

- for at least three years, and
- with a cadence of one measurement every 1 to 3 months.

In Figure 16 we show the distributions of standard deviations σ_{m_l} , σ_{m_c} and σ_χ at 4.85 GHz and 8.35 GHz. The sources exhibit a broad range of variability in all polarization parameters. The threshold for our search was set to the 20th percentile, P_{20} , i.e. one fifth of the corresponding standard deviation distribution, marked by the dotted lines in those plots.

In Tables 6 and 7, we list the sources with the most stable m_l and χ values at 4.85 GHz and 8.35 GHz. The names of sources for which both m_l and χ were found to be stable at at least one observing frequency are marked in bold face. The reported sources, exhibited significant linear polarization at least 95 % of the times they were observed. Significant linear polarization measurements are considered those with weighted mean of m_l at least three times larger than the weighted standard deviation in the corresponding session. In Table 8 we list linearly unpolarized sources, that is sources with no significant linear polarization measurements. For the latter, we report the average values of their m_l 3σ upper limits.

Finally, in Table 9, we list the sources with the most stable m_c . For comparison, we list the average and standard deviation values of m_c , corrected with both methods A and B as described in Sect. 3.7.1. Most of the sources in Table 9 exhibit mean m_c values very close to zero and therefore are considered circularly unpolarized. However, 3C 286, 3C 295, 3C 48 and CTA 102, show

significant circular polarization ($\langle m_c \rangle / \sigma_{m_c} \geq 3$) at 4.85 GHz with at least one of the two correction methods.

Komesaroff et al. (1984) observed two of the sources presented in Table 9 between December 1976 and March 1982, namely CTA 102 and PKS 1127-14. At that time CTA 102 showed variable circular polarization degree, which suggests that its m_c cannot be considered stable over such long time scales. On the other hand, PKS 1127-14 was found stable in m_c , with a synchronous decrease in both Stokes I and V . The average circular polarization degree of PKS 1127-14 over that period was $m_c \approx -0.1 \pm 0.03$, which is very close to the value listed in Table 9. This finding places additional bounds on its m_c stability, suggesting that it may have remained unchanged for ~ 40 years.

The sources reported in Tables 6–9 show stable behavior in different polarization properties. Nevertheless there is a small subgroup, namely 3C 286, 3C 295, 3C 48 and NGC 7027, which remain stable in both linear and circular polarization throughout the period we examined (2010.5–2016.3). Perley & Butler (2013a) and Zijlstra et al. (2008) show that these sources exhibit stable or well-predicted behavior also in Stokes I . Therefore they are best suited for the simultaneous calibration of all Stokes parameters.

5.1. Long-term stability of the circular polarization handedness

Previous studies have revealed a general tendency of the circular polarization handedness to remain stable over many years (e.g. Komesaroff et al. 1984; Homan & Wardle 1999; Homan et al. 2001). The consistency of the circular polarization sign may indicate either a consistent underlying ordered jet magnetic field component (e.g. toroidal or helical) or a general property of AGN with the sign of circular polarization set by the SMBH/accretion disk system (e.g. Enßlin 2003).

We compared our dataset with the one presented several decades ago in Komesaroff et al. (1984) to investigate the long-term stability of the circular polarization handedness, independently of the stability in the amplitude of m_c . There are ten common sources between the two samples. Three of them – namely PKS 1127-14, 3C 273 and 3C 279 – show stable circular polarization handedness in both datasets with the same sign of m_c . 3C 161 shows stable circular polarization handedness in both datasets but with the sign of m_c reversed. The rest of the common sources – namely PKS 0235+164, OJ 287, PKS 1510-089, PKS 1730-130, CTA 102 and 3C 454.3 – show short term variability of the circular polarization handedness in at least one of the two datasets. A thorough presentation of our circular polarization dataset and its comparison with previous works will be presented in a future publication.

6. Discussion and Conclusions

We presented the analysis of the radio linear and circular polarization of more than 150 sources observed with the Effelsberg 100-m telescope at 4.85 GHz and 8.35 GHz. The observations cover the period from July 2010 to April 2016 with a median cadence of around 1.2 months. We developed a new methodology for recovering all four Stokes parameters from the Effelsberg telescope observables. Although our method has been implemented for an observing system with circularly polarized feeds, it is easily generalizable to systems with linearly polarized feeds.

The novelty of our approach relies chiefly on the thorough treatment of the instrumental effects. In fact, our method aims

Table 5: The KS-test results (KS statistic, D , and p -value) for the comparison between the $\sigma_{I,Q,U}$ distributions for the 4.85 GHz data calibrated by the Müller method and our methodology. The results are presented for three polarized flux ranges: all, high polarization (≥ 100 mJy) and low polarization (15 mJy – 100 mJy). In column (4) we list the significance level at which the KS test null hypothesis can be rejected and in columns (5) and (6) we provide the median values of the corresponding $\sigma_{I,Q,U}$ distributions

Stokes	D	p	Significance level	median $\sigma_{I,Q,U}$ [Müller method] (mJy)	median $\sigma_{I,Q,U}$ [this work] (mJy)
All					
I	0.05	1.6×10^{-1}	1.4σ	11.829 ± 0.149	12.780 ± 0.074
* Q	0.15	1.2×10^{-10}	6.4σ	1.619 ± 0.007	1.467 ± 0.007
* U	0.13	4.7×10^{-8}	5.5σ	1.431 ± 0.008	1.324 ± 0.006
High polarization					
I	0.06	2.8×10^{-1}	1.1σ	24.911 ± 0.161	22.988 ± 0.105
Q	0.06	2.8×10^{-1}	1.1σ	2.376 ± 0.007	2.357 ± 0.007
U	0.08	8.8×10^{-2}	1.7σ	2.322 ± 0.008	2.247 ± 0.008
Low polarization					
I	0.09	2.4×10^{-2}	2.3σ	6.417 ± 0.132	7.379 ± 0.019
* Q	0.24	1.3×10^{-14}	7.7σ	1.099 ± 0.006	0.730 ± 0.004
* U	0.20	4.7×10^{-10}	6.2σ	1.009 ± 0.006	0.776 ± 0.004

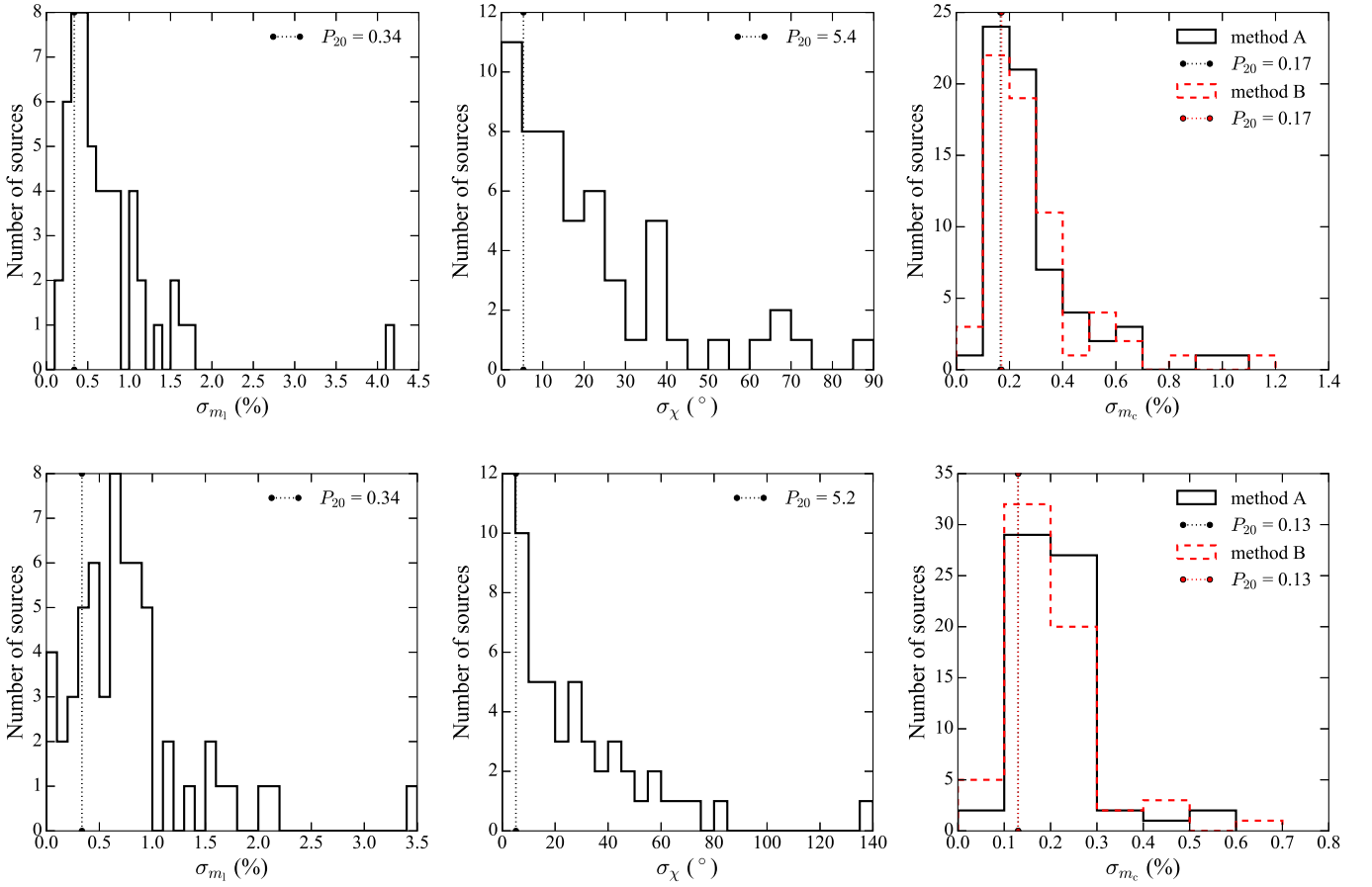


Fig. 16: Standard deviation distributions of the linear polarization degree, σ_{m_1} , polarization angle, σ_χ , and circular polarization degree, σ_{m_c} , measurements at 4.85 GHz (top row) and 8.35 GHz (lower row). The 20th percentile, P_{20} , of each standard deviation distribution is marked by a dotted line. The circular polarization data contain the measurements corrected with both methods A (continuous, black line) and B (dashed, red line) as described in Sect. 3.7.1.

Table 6: Sources with stable linear polarization degree, m_l . For each entry the upper row corresponds to 4.85 GHz and the lower to 8.35 GHz. The rows corresponding to frequencies at which the source was found to be either unpolarized or variable are filled with “...”. Sources with both stable m_l and χ are marked in bold face.

Source name	Δt_{obs} (yrs)	N_{obs}	N_{m_l}	$\langle I \rangle$ (Jy)	$\langle m_l \rangle$ (%)	σ_{m_l} (%)
3C 286	5.7	78	78	7.42	11.26	0.22
	5.7	75	75	5.11	11.88	0.07
3C 295
	5.5	47	47	3.37	0.93	0.07
3C 48	5.6	68	68	5.47	4.24	0.13
	5.6	69	69	3.20	5.61	0.07
NRAO 150	4.3	32	32	7.45	1.71	0.16
	4.3	33	33	9.14	1.27	0.22
MKN 501
	4.4	43	41	1.29	1.63	0.33

Notes. The entry in each column is as follows: (1) The source survey name, (2) the period that the source was observed, (3) the number of sessions that the source was observed, (4) the number of sessions that significant m_l was detected, (5) the average value of Stokes I , (6) the average value of m_l and (7) the standard deviation of m_l over the entire data set.

Table 7: Sources with stable polarization angle, χ . For each entry the upper row corresponds to 4.85 GHz and the lower to 8.35 GHz. The rows corresponding to frequencies at which the source was found to be either unpolarized or variable are filled with “...”. Sources with both stable m_l and χ are marked in bold face.

Source name	Δt_{obs} (yrs)	N_{obs}	N_χ	$\langle I \rangle$ (Jy)	$\langle \chi \rangle$ ($^\circ$)	σ_χ ($^\circ$)
3C 286	5.7	78	78	7.42	32.16	0.52
	5.7	75	75	5.11	33.07	0.23
3C 295
	5.5	47	47	3.37	31.90	3.68
3C 48	5.6	68	68	5.47	-73.65	0.98
	5.6	69	69	3.20	-64.25	0.32
PKS 0528+134	4.0	36	36	1.75	-26.72	2.69

S5 0836+71	4.0	33	32	2.46	-79.53	1.39
	4.0	31	31	2.57	-83.15	2.27
PKS 1127-14	4.0	38	37	3.10	-27.29	2.34
	4.0	38	38	2.71	-28.96	2.44
3C 273	4.3	52	52	35.84	-26.90	1.66
	5.2	51	51	27.09	-38.26	0.93
3C 454.3	5.5	52	52	11.70	4.01	2.03

Notes. The entry in each column is as follows: (1) The source survey name, (2) the period that the source was observed, (3) the number of sessions that the source was observed, (4) the number of sessions that significant m_l was detected, (5) the average value of Stokes I , (6) the average value of χ and (7) the standard deviation of χ over the entire data set.

Table 8: Linearly unpolarized sources. For each entry the upper row corresponds to 4.85 GHz and the lower to 8.35 GHz. The rows corresponding to frequencies at which the source was found to be polarized or variable are filled with “...”.

Source name	Δt_{obs} (yrs)	N_{obs}	$\langle I \rangle$ (Jy)	m_l (%)
3C 295	5.5	49	6.54	< 0.12

NGC 1052	4.0	36	1.24	< 1.10
	4.0	36	1.41	< 0.56
NGC 7027	5.3	53	5.37	< 0.16
	5.3	49	5.75	< 0.14

Notes. The entry in each column is as follows: (1) The source survey name, (2) the period that the source was observed, (3) the number of sessions that the source was observed, (4) the average value of Stokes I and (5) the upper limit of m_l over the entire data set.

at correcting the observables already prior to the computation of the Stokes parameters. In contrast, conventional methodologies – like the Müller matrix – operate on the Stokes vector. Consequently, cases of instrumentally corrupted observables that would be conventionally unusable, can be recovered by the careful treatment of their raw data. Additionally, for the correction of the circular polarization the Müller matrix method requires a good coverage of the parameter space. Our method on the other hand requires a small number of stable reference sources with the explicit knowledge of m_c only for one of them. Finally, the Müller matrix method lacks the capacity to treat the shortest observation cycle (sub-scan) operating on mean values.

In our method, the instrumental linear polarization – which is most likely caused by the slight ellipticity of the circular feed response – is modeled across the whole beam on the basis of the telescope response to unpolarized sources. Each sub-scan is then cleaned of the instrumental contribution, separately. Our results indicate that the instrumental linear polarization of the systems we used remained fairly stable throughout the period of 5.5 years we examined.

For the treatment of the instrumental circular polarization we introduced two independent methods: the zero-leveling of the m_c and the powerful singular value decomposition (SVD) method. Both rely on very few and easy to satisfy requirements and give very similar results. The results of both methods are also in agreement with the UMRAO data set.

The clean data are then subjected to a series of operations including the opacity and elevation-gain corrections which we found to be immune to the incident radiation’s polarization state. Moreover, the Airy disk beam pattern delivers amplitude estimates precise enough to accommodate reliable low circular polarization measurements.

All in all, our methodology allows us to minimize the uncertainties in linear and circular polarization degree at the level of 0.1–0.2 %. The polarization angle can be measured with an accuracy of the order of 1° .

We have estimated the instrumental rotation potentially caused by our apparatus by observing the Moon which has a simple radial configuration of the polarization angle. It provides then an excellent reference for (a) estimating the instrumental rotation and (b) conducting absolute angle calibration. We found that the instrument introduces a minute rotation of 1.26° and -0.5° for 4.85 GHz and 8.35 GHz, respectively. What is however worth noting is that we found evidence that there must be

Table 9: Sources with stable circular polarization degree, m_c . For each entry the upper row corresponds to 4.85 GHz and the lower to 8.35 GHz. The rows corresponding to frequencies at which the source was found to be variable are filled with “...”.

Source name	Δt_{obs} (yrs)	N_{obs}	$\langle I \rangle$ (Jy)	$N_{m_{c,A}}$	$\langle m_{c,A} \rangle$ (%)	$\sigma_{m_{c,A}}$ (%)	$N_{m_{c,B}}$	$\langle m_{c,B} \rangle$ (%)	$\sigma_{m_{c,B}}$ (%)
3C 286	5.7	78	7.42	53	-0.26	0.17	35	-0.36	0.09
	5.7	75	5.11	61	-0.11	0.12	39	-0.13	0.09
3C 295	5.5	49	6.54	24	-0.57	0.13	16	-0.68	0.11
	5.5	47	3.37	32	-0.16	0.08	20	-0.19	0.08
3C 48	5.6	68	5.47	50	-0.49	0.17	36	-0.60	0.09
	5.6	69	3.20	56	-0.12	0.12	39	-0.15	0.08
B2 0218+35
	3.2	20	1.26	18	0.04	0.12	13	0.02	0.07
4C +28.07
	4.2	40	3.20	38	0.14	0.12	23	0.15	0.10
NRAO 150
	4.3	33	9.14	33	-0.19	0.09	21	-0.21	0.11
PKS 1127-14	4.0	38	3.10	38	-0.20	0.17	29	-0.27	0.17

3C 345
	4.0	36	6.54	35	0.05	0.10	21	0.04	0.13
MKN 501	4.4	43	1.44	42	0.02	0.14	27	0.10	0.15
	4.4	43	1.29	42	0.04	0.12	28	0.02	0.12
CTA 102	4.3	44	4.57	44	-0.23	0.14	31	-0.32	0.08

NGC 7027	5.3	53	5.37	39	0.10	0.08	36	0.06	0.11
	5.3	49	5.75	49	0.03	0.05	39	0.01	0.08

Notes. The entry in each column is as follows: (1) The source survey name, (2) the period that the source was observed, (3) the number of sessions that the source was observed, (4) the average value of Stokes I over the entire data set, (5) the number of sessions that the m_c measurements were corrected with method A, (6) the average value of m_c , corrected with method A, (7) the standard deviation of m_c , corrected with method A, (8) the number of sessions that the m_c measurements were corrected with method B, (9) the average value of m_c , corrected with method B, (10) the standard deviation of m_c , corrected with method B, over the entire data set.

an, at least, mild dependence of the rotation on the source elevation. For completeness further investigation is worthy despite the marginal magnitude of the effect.

Despite the conceptual differences between our method and the Müller matrix one, we conducted a quantitative comparison of their effectiveness. We examined the intra-session variability in the linear polarization parameters. Those should remain unchanged over such short time scales even for intrinsically variable sources. Our methodology performs significantly better, particularly for low linear polarization observations (15–100 mJy polarized flux), where it delivers 28 % more stable Stokes Q and U results than the Müller method.

After having reconstructed as accurately as possible the polarization state of our sample, we searched for sources with stable polarization characteristics. We found five sources with significant and stable linear polarization. A list of three sources remain constantly unpolarized over the entire period of almost 5.5 years we examined. A total of 11 sources were found to have stable circular polarization degree four of which with non-zero m_c . One of the sources with stable circular polarization degree, namely PKS 1127-14, was found to be stable and at the same level several decades ago by Komesaroff et al. (1984), suggesting that its m_c may have remained unchanged for ~ 40 years. Additionally, we found eight sources that maintain a stable polarization angle over the examined period. All this is provided to the community for future polarization observations reference.

Finally, we investigated the long-term stability of the circular polarization handedness for the ten common sources between our sample and the dataset presented in Komesaroff et al. (1984).

Three sources show stable circular polarization handedness in both datasets with the same sign of m_c , one with the opposite sign and the other six sources show short term variability of the circular polarization handedness in at least one of the two datasets.

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Appendix A: Instrument model for the 4.85 GHz and 8.35 GHz Effelsberg receivers

The functional forms of the instrument models – one for each Stokes parameter and scanning direction – for the 4.85 GHz receiver are:

$$M_{Q,azi} = \alpha_1 I e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} \quad (\text{A.1})$$

$$M_{Q,elv} = \alpha_1 I e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} + \alpha_3 I e^{-\frac{[x-(\mu-\beta_3)]^2}{2(\gamma_3\sigma)^2}} \quad (\text{A.2})$$

$$M_{U,azi} = \frac{\alpha_1 I [x - (\mu - \beta_1)] e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}}}{(\gamma_1\sigma)^2} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} \quad (\text{A.3})$$

$$M_{U,elv} = \frac{\alpha_1 I [x - (\mu - \beta_1)] e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}}}{(\gamma_1\sigma)^2} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} \quad (\text{A.4})$$

and for the 8.35 GHz receiver:

$$M_{Q,azi} = \alpha_1 I e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} + \alpha_3 I e^{-\frac{[x-(\mu-\beta_3)]^2}{2(\gamma_3\sigma)^2}} \quad (\text{A.5})$$

$$M_{Q,elv} = \alpha_1 I e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} \quad (\text{A.6})$$

$$M_{U,azi} = \alpha_1 I e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}} \quad (\text{A.7})$$

$$M_{U,elv} = \frac{\alpha_1 I [x - (\mu - \beta_1)] e^{-\frac{[x-(\mu-\beta_1)]^2}{2(\gamma_1\sigma)^2}}}{(\gamma_1\sigma)^2} + \alpha_2 I e^{-\frac{[x-(\mu-\beta_2)]^2}{2(\gamma_2\sigma)^2}} \quad (\text{A.8})$$

where,

- azi, elv the scanning direction
- I is the measured mean amplitude of the LCP and RCP signals
- μ is the measured mean offset of the LCP and RCP signals
- σ is the measured mean FWHM of the LCP and RCP signals
- $\alpha_j, \beta_j, \gamma_j$ are the fitted parameters for each model with $j = 1, 2, 3$

In Table A.1 we show a set of initial parameter values that we use as starting point for the fitting algorithm.

Appendix B: Feed ellipticity and the measurement of Stokes parameters

In this appendix we estimate the effect of feed ellipticity on the measurement of Stokes parameters. In the following, we provide an elementary approach where several aspects have been oversimplified, e.g. parameters a and b in Eqs. B.1 and B.2 are considered real instead of complex numbers. This approach was selected in order to derive a rough estimate of the effect on the measured parameters. A thorough study of the effect can be found in e.g. McKinnon (1992) or Cenacchi et al. (2009).

The feed ellipticity can be parameterized as a cross-talk between the left- and right-hand circularly polarized electric field components recorded by the system:

$$E'_L(t) = E_L(t) + aE_R(t) = E_L e^{i\omega t} + aE_R e^{i(\omega t + \delta)}, \quad (\text{B.1})$$

$$E'_R(t) = E_R(t) + bE_L(t) = E_R e^{i(\omega t + \delta)} + bE_L e^{i\omega t}, \quad (\text{B.2})$$

where,

- $E_{L,R}$ the amplitudes of the two incident (orthogonal) circularly polarized electric field components
- ω the angular frequency of the electromagnetic wave
- δ the phase difference between $E_L(t)$ and $E_R(t)$
- a, b the percentage of the incident left-hand polarized electric field component recorded by the right-hand polarized channel and vice versa.

The primed and unprimed quantities in Eqs. B.1 and B.2 refer to the recorded and incident signals, respectively. The terms which contain parameters a and b appear due to the ellipticity of the circular feed response (see Eqs. 5 and 6 for comparison).

Using Eqs. B.1 and B.2, the LCP, RCP, COS and SIN signals can be written as:

$$\text{LCP} = \langle E'_L E'_L \rangle = \langle E_L^2 + aE_L E_R e^{i\delta} + aE_L E_R e^{-i\delta} + a^2 E_R^2 \rangle, \quad (\text{B.3})$$

$$\text{RCP} = \langle E'_R E'_R \rangle = \langle E_R^2 + bE_L E_R e^{-i\delta} + bE_L E_R e^{i\delta} + b^2 E_L^2 \rangle, \quad (\text{B.4})$$

$$\text{COS} = \langle E'_L E'_R \rangle = \langle E_L E_R e^{i\delta} + bE_L^2 + aE_R^2 + abE_L E_R e^{-i\delta} \rangle, \quad (\text{B.5})$$

$$\text{SIN} = \langle E'_L E'_R \rangle_{90^\circ} = \langle E_L E_R e^{i(\delta-90^\circ)} + bE_L^2 + aE_R^2 + abE_L E_R e^{-i(\delta-90^\circ)} \rangle, \quad (\text{B.6})$$

where the “*” denotes the complex conjugate and the subscript “90°” of Eq. B.6 denotes that an additional phase difference of 90° is introduced between $E'_L(t)$ and $E'_R(t)$. The real part of Eqs. B.3–B.6, which is recorded by the system, is:

$$\text{LCP} = \langle E_L^2 + a^2 E_R^2 + 2aE_L E_R \cos \delta \rangle, \quad (\text{B.7})$$

$$\text{RCP} = \langle E_R^2 + b^2 E_L^2 + 2bE_L E_R \cos \delta \rangle, \quad (\text{B.8})$$

$$\text{COS} = \langle (ab + 1)E_L E_R \cos \delta + aE_R^2 + bE_L^2 \rangle, \quad (\text{B.9})$$

$$\text{SIN} = \langle (ab + 1)E_L E_R \sin \delta + aE_R^2 + bE_L^2 \rangle, \quad (\text{B.10})$$

Table A.1: The initial values of the Stokes Q and U instrument model parameters for the 4.85 GHz and 8.35 GHz receivers. The functional forms of the model are given in Eqs.A.1–A.8.

Model	ν_{obs} (GHz)	α_1	β_1	γ_1	α_2	β_2	γ_2	α_3	β_3	γ_3
$M_{Q,\text{azi}}$	4.85	-0.005	-70	0.5	-0.002	-5	0.5
$M_{Q,\text{elv}}$	4.85	0.003	110	0.2	-0.010	0	0.2	0.002	-110	0.2
$M_{U,\text{azi}}$	4.85	0.3	0	0.9	-1	40	0.5
$M_{U,\text{elv}}$	4.85	-0.3	5	0.3	-0.5	0	0.2
$M_{Q,\text{azi}}$	8.35	0.001	67	0.2	-0.003	-6	0.3	0.001	-77	0.2
$M_{Q,\text{elv}}$	8.35	-0.002	22	0.4	-0.002	-35	0.3
$M_{U,\text{azi}}$	8.35	0.005	-5	0.4
$M_{U,\text{elv}}$	8.35	0.1	-8	0.9	0.002	44	0.3

To derive a rough estimate of the effect on the measurement of the four Stokes parameters, we can further simplify the above expressions by assuming that $a \approx b = k$:

$$\text{LCP} = \langle E_L^2 + k^2 E_R^2 + 2kE_L E_R \cos \delta \rangle, \quad (\text{B.11})$$

$$\text{RCP} = \langle E_R^2 + k^2 E_L^2 + 2kE_L E_R \cos \delta \rangle, \quad (\text{B.12})$$

$$\text{COS} = \langle (k^2 + 1)E_L E_R \cos \delta + k(E_R^2 + E_L^2) \rangle, \quad (\text{B.13})$$

$$\text{SIN} = \langle (k^2 + 1)E_L E_R \sin \delta + k(E_R^2 + E_L^2) \rangle, \quad (\text{B.14})$$

The last terms in Eqs. B.13 and B.14 describe a contribution of Stokes I ($= \langle E_R^2 \rangle + \langle E_L^2 \rangle$) to the measured linearly polarized flux density as recorded by the COS and SIN channels. We can estimate the parameter k from the observations of linearly unpolarized sources. Such sources have a random phase difference δ , which means that $\langle \cos \delta \rangle = 0$ and hence the first terms of Eqs. B.13 and B.14 are vanished. The last terms, on the other hand, describe the spurious instrumental linear polarization signals that we correct for using the instrument model (see Sect. 3.2). Thus the parameter k is described (across the whole beam) by the instrument model (e.g. Fig. 4) which is at maximum 0.005 for the systems we used. Therefore – in this simplified approach – we can exclude the second order terms of k from Eqs. B.11–B.14 since $k^2 \rightarrow 0$ (it is at maximum $2.5 \cdot 10^{-5}$) which results in:

$$\text{LCP} = \langle E_L^2 + 2kE_L E_R \cos \delta \rangle, \quad (\text{B.15})$$

$$\text{RCP} = \langle E_R^2 + 2kE_L E_R \cos \delta \rangle, \quad (\text{B.16})$$

$$\text{COS} = \langle E_L E_R \cos \delta + k(E_R^2 + E_L^2) \rangle, \quad (\text{B.17})$$

$$\text{SIN} = \langle E_L E_R \sin \delta + k(E_R^2 + E_L^2) \rangle, \quad (\text{B.18})$$

Equations B.15–B.18 can be used to estimate the effect of the feed ellipticity to the measurement of the Stokes parameters. The last terms in Eqs. B.17 and B.18 are removed by the instrumental linear polarization correction scheme as described in Sect. 3.2. Therefore, Stokes Q and U should not be affected by the feed ellipticity after this correction step. Stokes V remains also unaffected because the subtraction of LCP from RCP removes the last terms of Eqs. B.15 and B.16 which appear due to the feed ellipticity effect.

Finally, Stokes I seems to be affected by the feed ellipticity effect even in this elementary approach. The recorded Stokes I can deviate from the incident one by $4kE_L E_R \cos \delta$, which is proportional to the linear polarization of the source ($E_L E_R \cos \delta = \frac{Q}{2}$). Using the average linearly polarized flux density over all our

measurements (0.125 Jy) as an estimate of Stokes Q in the above expression we get:

$$4kE_L E_R \cos \delta = 4k \frac{Q}{2} \approx 4 \cdot 0.005 \cdot \frac{0.125}{2} = 0.00125 \text{ Jy} \quad (\text{B.19})$$

Therefore, the measured Stokes I deviates from the incident one by ~ 1 mJy on average. This value is much smaller than the average uncertainty of our measurements (15–20 mJy, Table 2). In fact, if we use the maximum linearly polarized flux density that we have ever measured (2.116 Jy) we calculate that the feed ellipticity effect on Stokes I is at maximum 21 mJy, which is comparable to the average uncertainty.

The above discussion shows that, in most cases, the feed ellipticity doesn't seem to have any measurable effect on the Stokes parameter measurements once we account for the instrumental linear polarization (e.g. using the correction scheme presented in Sect. 3.2). The parameter which is mostly affected is Stokes I but the effect becomes significant only when the linearly polarized flux density of the observed source is particularly large, e.g. ≥ 2 Jy.

Appendix C: Circular polarization curves

In Fig. C.1 we show the circular polarization degree curves for five common sources with the UMRAO monitoring program. The two datasets are in very good agreement as described in Sect. 3.7.2.

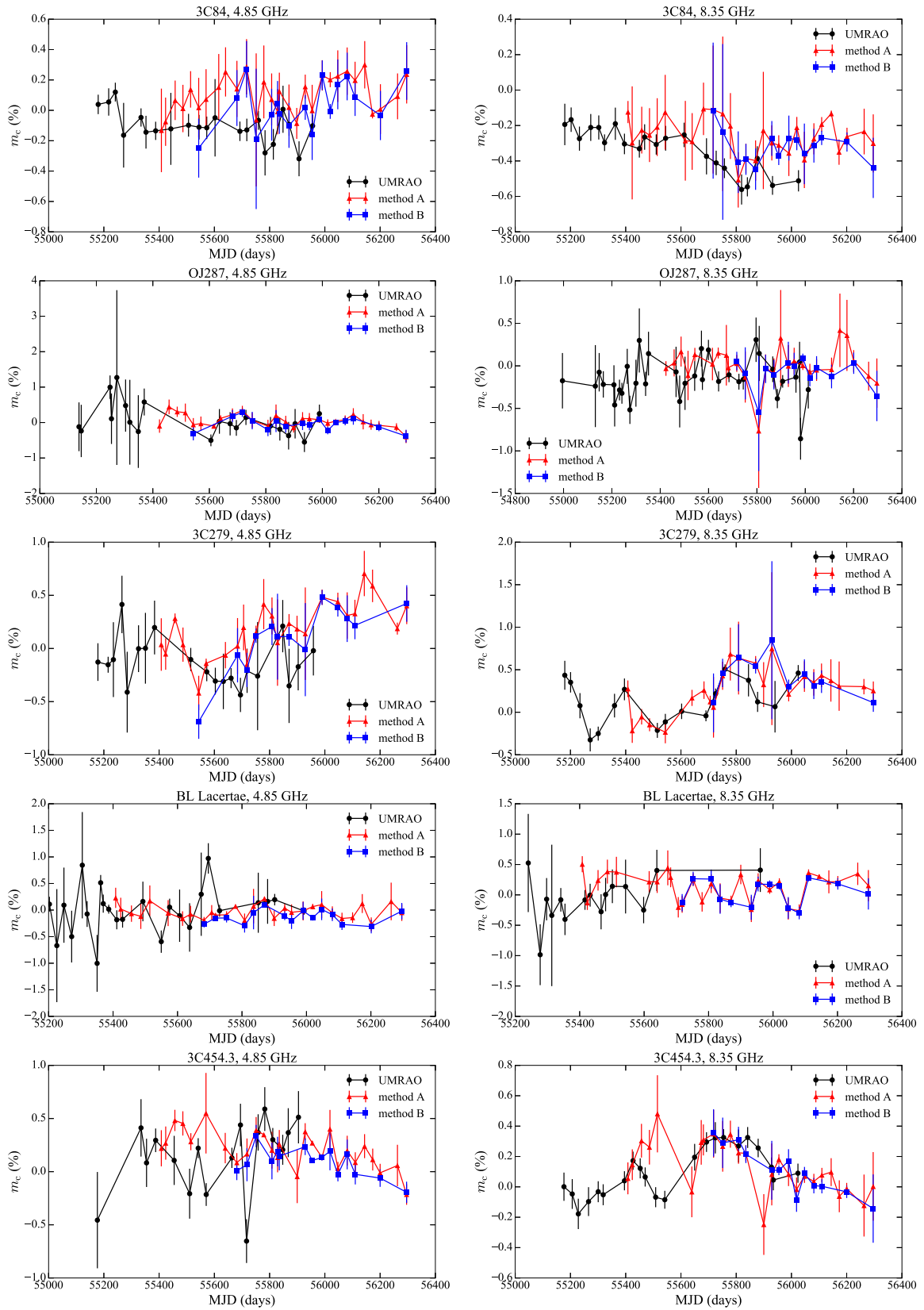


Fig. C.1: Circular polarization curves of the UMRao dataset over-plotted against our results using both calibration methods A and B as described in Sect. 3.7. The comparison is performed over five sources with overlapping data sets, which were observed at 4.85 GHz (left column) and 8.35 GHz (right column).